

NeurIPS2021 Tutorial

Real-Time Optimization for Fast and Complex Control Systems

Part 1

Introduction to Control Systems



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Goal of This Tutorial

- To help researchers and engineers in the field of machine learning tackle problems in **control systems**
 - Control systems involve **real-time decision making**: a kind of artificial intelligence
 - Overview of **control theory** that may be helpful for proper use of machine learning
 - Primary focus: **model predictive control (MPC)** based on real-time optimization. MPC can address various control problems beyond such traditional control objectives as regulation and tracking.

Outline

Part 1: Introduction to Control Systems

Part 2: Optimal Control and Model Predictive Control

Part 3: Real-Time Optimization for Model Predictive Control

Part 4: Advanced Topics in Model Predictive Control

Outline of Part 1

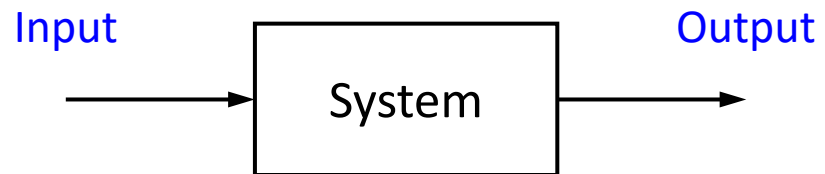
- What is control system?
- Concepts and methods for analysis and design of control systems
 - Mathematical models, modeling, identification, stability, etc.
 - Optimal control, adaptive/learning control, robust control, etc.

What is Control?

- To operate a **system** as desired

What is System?

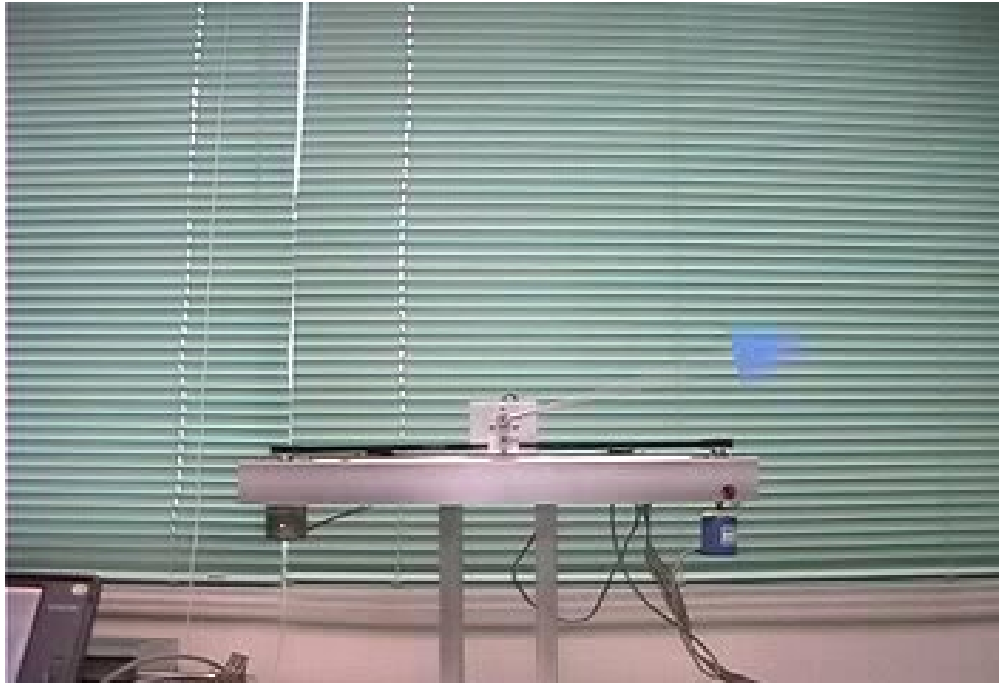
- Something changing dynamically according to inputs



Block Diagram

Input and Output:
Signals (Functions of Time)

Control Systems



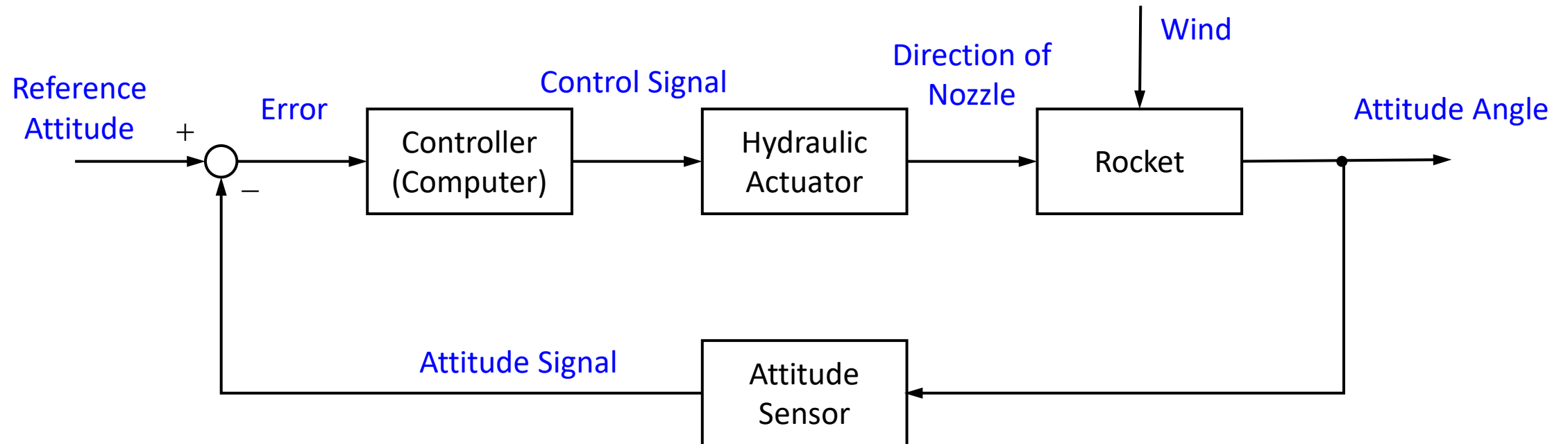
Inverted Pendulum © Toru Asai 2004



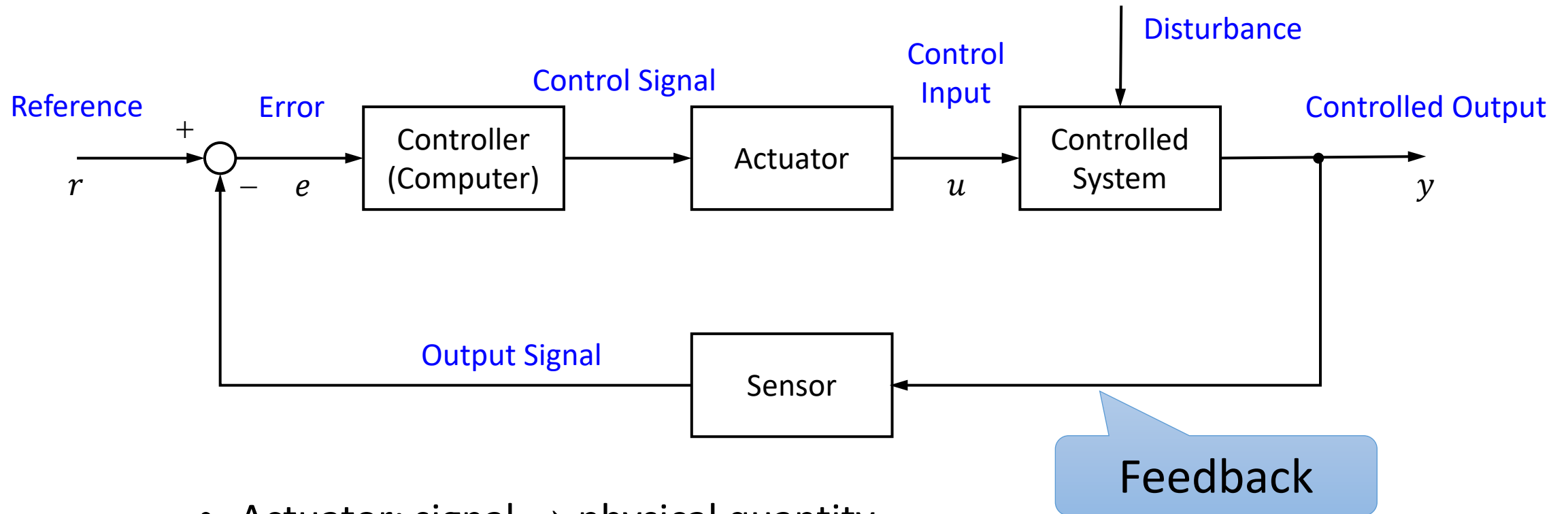
Rocket © JAXA 2014

Systems kept upward by control against gravity

Attitude Control System of a Rocket

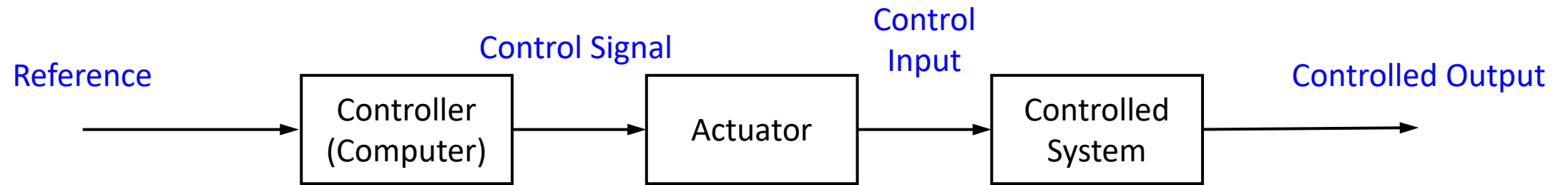


Feedback Control System (Closed-Loop)



- Actuator: signal \rightarrow physical quantity
- Sensor: physical quantity \rightarrow signal
- Actuator/sensor blocks are often omitted.

Feedforward Control System (Open-Loop)



- No sensor
- No disturbance

Control Systems are Everywhere

- Such machines as cars, ships, aircraft, and robots
 - **Inputs:** forces, torques, steering
 - **Outputs:** positions, velocities, directions
- Temperature, environment, economy, and epidemic
 - **Inputs:** heat, gas emissions, monetary policy, mask/vaccine mandate
 - **Outputs:** temperature, atmospheric constituent, money supply, spread rate

Control Engineering

- Methodology to **analyze** and **design** control systems
- Methodology based on mathematical models of control systems:
Control Theory
- A lot of definitions, **theorems** and proofs: Stability, Controllability, Optimality, etc.

Mathematical Models

- **System:** Mapping from input signal (function of time) to output signal

$$y = P(u), \quad P: \text{Mapping between function spaces}$$

- **Input-Output Model**

$$y^{(n)}(t) = F(y^{(n-1)}(t), \dots, \dot{y}(t), y(t), u^{(m)}(t), \dots, \dot{u}(t), u(t), t)$$

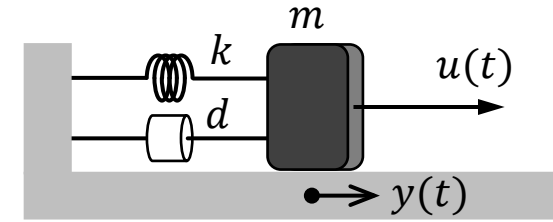
($\dot{}$) Time Derivative

- **State-Space Representation**

$$\begin{aligned} \dot{x}(t) &= f(x(t), u(t), t), & x(t): \text{Vector of state variables} \\ y(t) &= h(x(t), u(t), t) \end{aligned}$$

- **Continuous-valued signals**
- Continuous time / Discrete time: differential equations / difference equations
- Stochastic Systems: involves random variables
- Hybrid systems: mixture of continuous dynamics and discrete events

Example: Mass-Spring System



- Input-Output Model ($y(t)$: displacement, $u(t)$: external force)

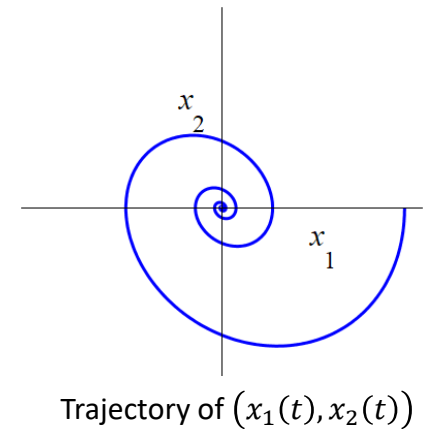
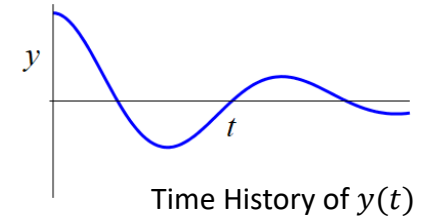
$$m\ddot{y}(t) + d\dot{y}(t) + ky(t) = u(t)$$

- State-Space Representation

($x_1(t)$: displacement, $x_2(t)$: velocity, $u(t)$: external force)

$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} x_2(t) \\ -\frac{k}{m}x_1(t) - \frac{d}{m}x_2(t) + \frac{1}{m}u(t) \end{bmatrix}$$

$$y(t) = x_1(t)$$



Linear Time-Invariant (LTI) Systems

- Input-Output Model (Single-Input Single-Output: SISO)

$$a_n y^{(n)}(t) + a_{n-1} y^{(n-1)}(t) + \dots + a_1 \dot{y}(t) + a_0 y(t) \\ = b_m u^{(m)}(t) + \dots + b_1 \dot{u}(t) + b_0 u(t)$$

- State-Space Representation (Multiple-Input Multiple-Output: MIMO)

$$\dot{x}(t) = Ax(t) + Bu(t) \quad A, B, C, D: \text{Matrices} \\ y(t) = Cx(t) + Du(t)$$

- Transfer Function $y(s) = P(s)u(s)$

$$y(s) = \frac{b_m s^m + \dots + b_1 s + b_0}{a_n s^n + \dots + a_1 s + a_0} u(s), \quad y(s) = [C(sI - A)^{-1}B + D]u(s)$$

Modeling/Identification

- **Modeling:** Construction of mathematical models based on knowledge
- Model Structures: LTI, Wiener, Hammerstein, Volterra
- Model Transformation: Order Reduction, Structure Simplification
- **Identification:** Construction of mathematical models from data
 - Parametric/Nonparametric
 - Prediction Error Method
 - Subspace Identification
 - Learning Dynamical Systems

L. Ljung: System Identification: Theory for the User, Prentice Hall (1998)

O. Nelles: Nonlinear System Identification, Springer (2001); S. A. Billings: Nonlinear System Identification, Wiley (2013)

K. Fujimoto, J. M. A. Scherpen: Balanced Realization and Model Order Reduction for Nonlinear Systems Based on Singular Value Analysis; SIAM J. Contr. and Optim., 48(7), 4591-4623 (2010)

T. Ohtsuka: Model Structure Simplification of Nonlinear Systems via Immersion; IEEE Trans. Autom. Contr., 50(5), 607-618 (2005)

Analysis

- **Stability:** Input-Output, Lyapunov, Input-to-State
- **Gain:** $\|y\| \leq \gamma \|u\| + \beta$, $\|\cdot\|$ norm of a signal
- Passivity, Dissipativity
- LTI: Frequency Response $G(j\omega)$ ($j = \sqrt{-1}$), Bode Plot, Vector Locus
- Controllability/Reachability (Existence of Input Signal for Given Initial/Terminal State)
- Observability (Uniqueness of Initial State for Given Output Signal)
- Invariance of a Set/Manifold (Unreachability, Safety Guarantee)

Stability Analysis

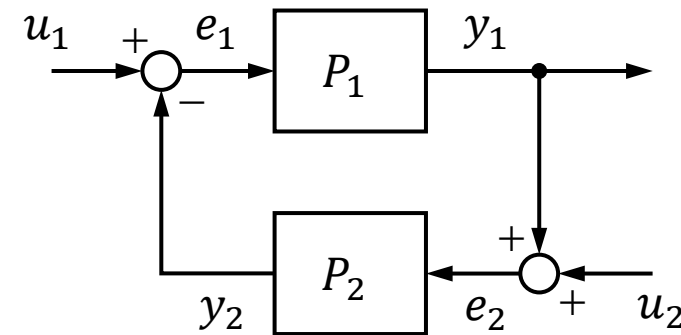
- LTI: Routh/Hurwitz Criterion, Nyquist Criterion, Eigenvalues
- **Lyapunov Function:** Let $x = 0$ be an equilibrium point of $\dot{x}(t) = f(x(t))$. If there is a continuously differentiable function $V(x)$ in a neighborhood D of $x = 0$ such that $V(0) = 0$, $V(x) > 0$ in $D - \{0\}$ and $\dot{V}(x(t)) < 0$ in $D - \{0\}$ then $x = 0$ is asymptotically stable.
- **Convex Optimization** to Find $V(x)$: Linear Matrix Inequalities (LMI), Sum-of-Squares (SOS) Programming

Stability can be checked without solving differential equations!

Stability Analysis

- **Small Gain Theorem:** Suppose two systems P_1 and P_2 have finite gains γ_1 and γ_2 . If $\gamma_1\gamma_2 < 1$ holds then their feedback connection also has a finite gain as a system with input (u_1, u_2) and output (e_1, e_2) .
- **Passivity Theorem:** If two systems P_1 and P_2 are passive then their feedback connection is also passive.

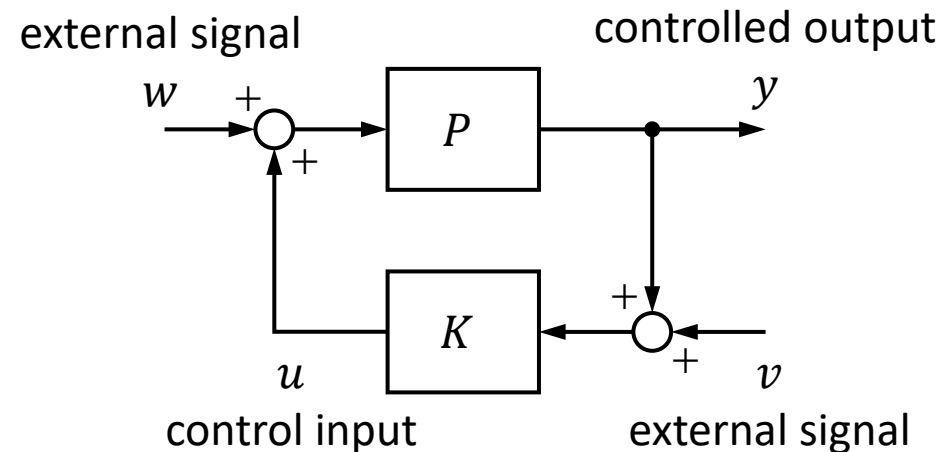
Stability can be guaranteed
without detailed models!



Feedback Connection

Control Design

- For a given system $y = P(u)$, find a controller (a system) $u = K(y)$ so that design specifications are satisfied.
- Not always but often formulated as a constrained optimization problem.



Feedback Control System

Control Design Methods

- PID (Proportional-Integral-Derivative), Loop Shaping
- **State Feedback** (+ State Estimation)
 - Pole Assignment, Control Lyapunov Function
 - Optimal Control
 - Sliding Mode Control
 - Feedback Linearization
- Adaptive Control, Iterative Learning Control
- Robust Control
- Distributed Control

Optimal Control

Find $u(t)$ (**feedforward**) or $u(x, t)$ (**state feedback**) ($0 \leq t \leq T$)

minimizing $J = \varphi(x(T), T) + \int_0^T L(x(t), u(t), t) dt$

subject to $\dot{x}(t) = f(x(t), u(t), t)$, $x(0)$ given

$$C(x(t), u(t), t) = 0$$

$$D(x(t), u(t), t) \leq 0$$

$$\psi(x(T), T) = 0, \quad \chi(x(T), T) \leq 0$$

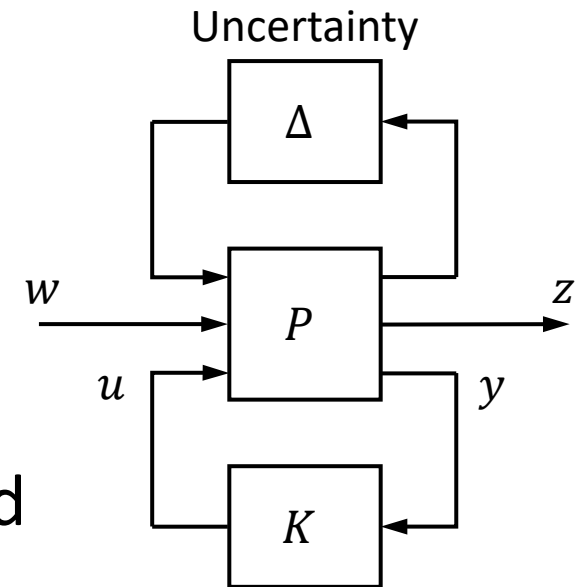
Terminal time T can be either given or free.

Adaptive Control, Iterative Learning Control

- **Adaptive Control:** Parameterized controller $u = K(y; \theta)$ and adaptation law to adjust θ
e.g.) on-line estimation of unknown parameter θ in the system model
- **Iterative Learning Control:** Iteratively update $u(t)$ ($0 \leq t \leq T$) to achieve perfect tracking based only on tracking error $e(t)$ ($0 \leq t \leq T$) with almost no prior knowledge on the system
e.g.) $u_{k+1}(t) = u_k(t) + K\ddot{e}_k(t)$

Robust Control

- **Uncertainty:** System Δ in a unit ball $B = \{\Delta: \|\Delta\| < 1\}$
- Δ is uncertain but deterministic
- **Robust Stabilization:** Find a controller K such that the closed-loop system from w to z is stable for any $\Delta \in B$.
- **Robust Performance:** Find a controller K such that performance specifications from w to z are satisfied for any $\Delta \in B$.
- Basic Tool: Small Gain Theorem



Feedback Control System
with Uncertainty

Distributed Control

- Find a controller K utilizing limited information

$$u_i = K_i(y_{j_{i1}}, \dots, y_{j_{im}})$$

- **Multi-agent system:** Global task (formation, consensus, coverage, etc.) by a set of systems (agents) with distributed control
- **Decentralized Control:** No communication between controllers

$$u_i = K_i(y_i)$$

Summary

- **Control systems** involve real-time decision making, a kind of artificial intelligence.
- Control systems are everywhere from machines to environment and society.
- **Control theory** provides mathematical tools for analysis and design of control systems.
- **Mathematical models** of systems play crucial roles in control theory. However, there are some nice methods to deal with qualitative characteristics and uncertainties without detailed models.

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Real-Time Optimization for Fast and Complex Control Systems

Part 2

Optimal Control and Model Predictive Control



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Outline

Part 1: Introduction to Control Systems

Part 2: Optimal Control and Model Predictive Control

Part 3: Real-Time Optimization for Model Predictive Control

Part 4: Advanced Topics in Model Predictive Control

Outline of Part 2

- Optimal control problem, Euler-Lagrange Equations (ELE), Hamilton-Jacobi-Bellman Equation (HJBE), and numerical solution methods
- Model Predictive Control (MPC): problem formulation and difficulties, computation and stability

Optimal Control Problem

Find $u(t)$ (feedforward) or $u(x, t)$ (state feedback) ($0 \leq t \leq T$)

minimizing $J = \varphi(x(T), T) + \int_0^T L(x(t), u(t), t) dt$

subject to $\dot{x}(t) = f(x(t), u(t), t)$, $x(0)$ given

We omit the constraints except for the state equation for simplicity. They can be handled by penalty functions or barrier functions. The terminal time T is assumed to be given.

Euler-Lagrange Equations (ELE): Stationary Conditions for Optimal Trajectory

$$\dot{x}(t) = f(x(t), u(t), t), \quad x(0) \text{ given}$$

$$\dot{\lambda}(t) = -\nabla_x H(x(t), u(t), \lambda(t), t), \quad \lambda(T) = \nabla_x \varphi(x(T), T)$$

$$\nabla_u H(x(t), u(t), \lambda(t), t) = 0$$

Nonlinear Two-Point Boundary-Value Problem (TPBVP)

where

$$H(x, u, \lambda, t) = L(x, u, t) + \lambda^T f(x, u, t) : \text{Hamiltonian}$$

$\lambda(t)$: costate or adjoint variable

Solution $u(t)$: Candidate of optimal feedforward control for given $x(0)$

Numerical Solution of ELE

- **Iterative search in the space of functions** $u(t)$ (single shooting) or $(x(t), \lambda(t), u(t))$ (multiple shooting)
- Gradient methods: Steepest descent, conjugate gradient method
- Newton's method
- Quasi-Newton method
- Some special structures are exploited
- Time horizon is normally discretized for numerical solution
- Computationally demanding!

Hamilton-Jacobi-Bellman Equation (HJBE)

Value Function

$$V(x, t) = \min_{u[t, T]} \left[\varphi(x(T)) + \int_t^T L(x(t), u(t), t) dt \right]$$

HJBE

$$-\nabla_t V(x, t) = \min_u H(x, u, \nabla_x V(x, t), t), \quad V(x, T) = \varphi(x, T)$$

Nonlinear Partial Differential Equation (PDE)

Optimal Control

$$u_{opt}(x, t) = \arg \min_u H(x, u, \nabla_x V(x, t), t)$$

State Feedback!

Solvable Example: Linear Quadratic Control

Linear system: $\dot{x}(t) = Ax(t) + Bu(t)$

Quadratic cost:

$$J = \frac{1}{2} x^T(T) S_f x(T) + \int_0^T \frac{1}{2} \left(x^T(t) Q x(t) + u^T(t) R u(t) \right) dt$$

Optimal control: $u_{opt}(x, t) = \underline{-R^{-1} B^T S(t) x}$

State Feedback with
a Time-Varying Gain Matrix

Riccati differential equation:

$$-\dot{S}(t) = A^T S(t) + S(t) A - S(t) B R^{-1} B^T S(t) + Q, \quad S(T) = S_f$$

Value function: $V(x, t) = \frac{1}{2} x^T S(t) x$

Costate in ELE: $\lambda(t) = S(t) x(t)$

Numerical Solution of HJBE

- If $V(x, t)$ can be stored for all (x, t) then HJBE can be solved numerically backward in time starting from $V(x, T) = \varphi(x)$
- However, impractical due to explosive growth of computation and storage for high dimensional systems: **curse of dimensionality**
- Approximate solution methods: power series (Al'Brekht '61; Lukes '69), Galerkin method (Beard et al. '97), interpolation (Kreiselmeier&Birkhölzer '94), neural networks (Goh '93), RBF (Huang et al. '06), GPR (Fujimoto et al. '18), etc.

Infinite-Horizon Problem

Stationary solution to infinite-horizon optimal control problem

$$\text{HJE: } H(x, u(x, \lambda), \lambda) = 0, \quad \lambda = \nabla_x V(x)$$

$$u(x, \lambda) = \arg \min_u H(x, u, \lambda)$$

ELE as a Hamiltonian System (Canonical Equations)

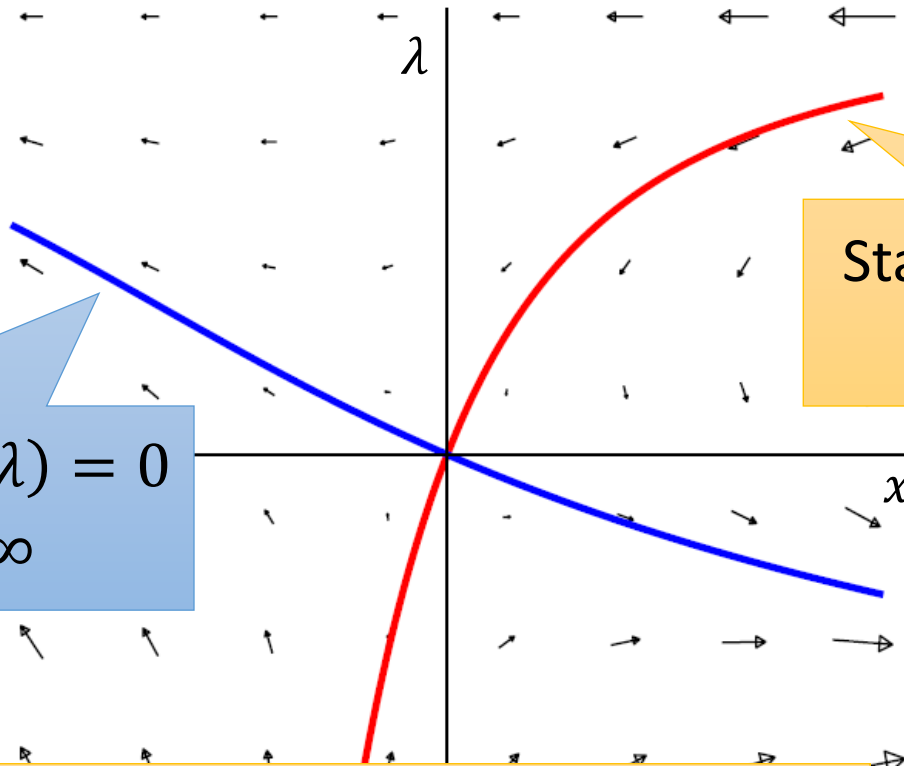
$$\dot{x}(t) = \nabla_\lambda H(x(t), \lambda(t))$$

$$\dot{\lambda}(t) = -\nabla_x H(x(t), \lambda(t))$$

where $H(x, \lambda) := H(x, u(x, \lambda), \lambda)$

Terminal condition for $\lambda(t)$ ($t \rightarrow \infty$)?

Stable Manifold in Infinite-Horizon Problem



Unstable Manifold: $H(x, \lambda) = 0$
and $(x(t), \lambda(t)) \rightarrow \infty$

Stable Manifold: $H(x, \lambda) = 0$
and $(x(t), \lambda(t)) \rightarrow 0$

Stabilizing solution to infinite-horizon problem
= **Stable manifold** of Hamiltonian system

Solution methods to
find stable manifold

(van der Schaft '91)

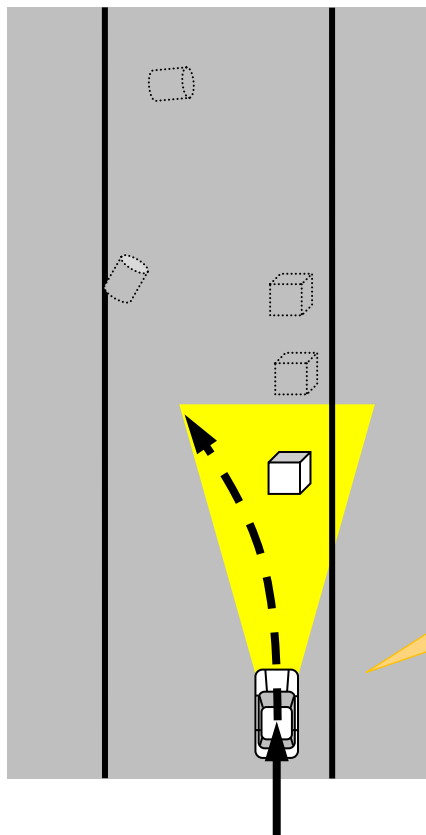
(Sakamoto&van der Schaft '08; O '11)

Outline of Part 2

- Optimal control problem, Euler-Lagrange Equations (ELE), Hamilton-Jacobi-Bellman Equation (HJBE), and numerical solution methods
- Model Predictive Control (MPC): problem formulation and difficulties, computation and stability

What is MPC?

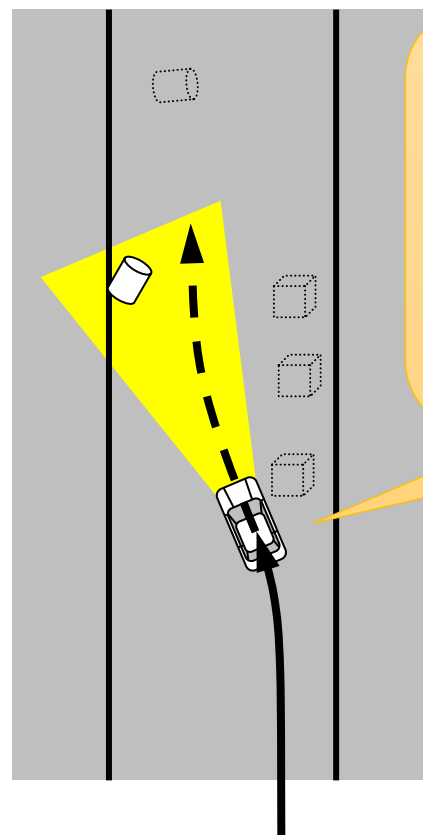
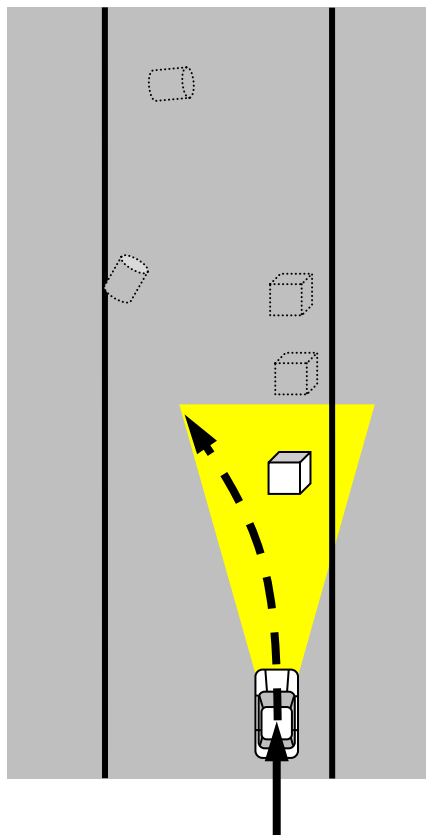
Feedback control by **real-time optimization** of the system response over a finite future.



Current control action is determined by optimization over **a foreseeable future**.

What is MPC?

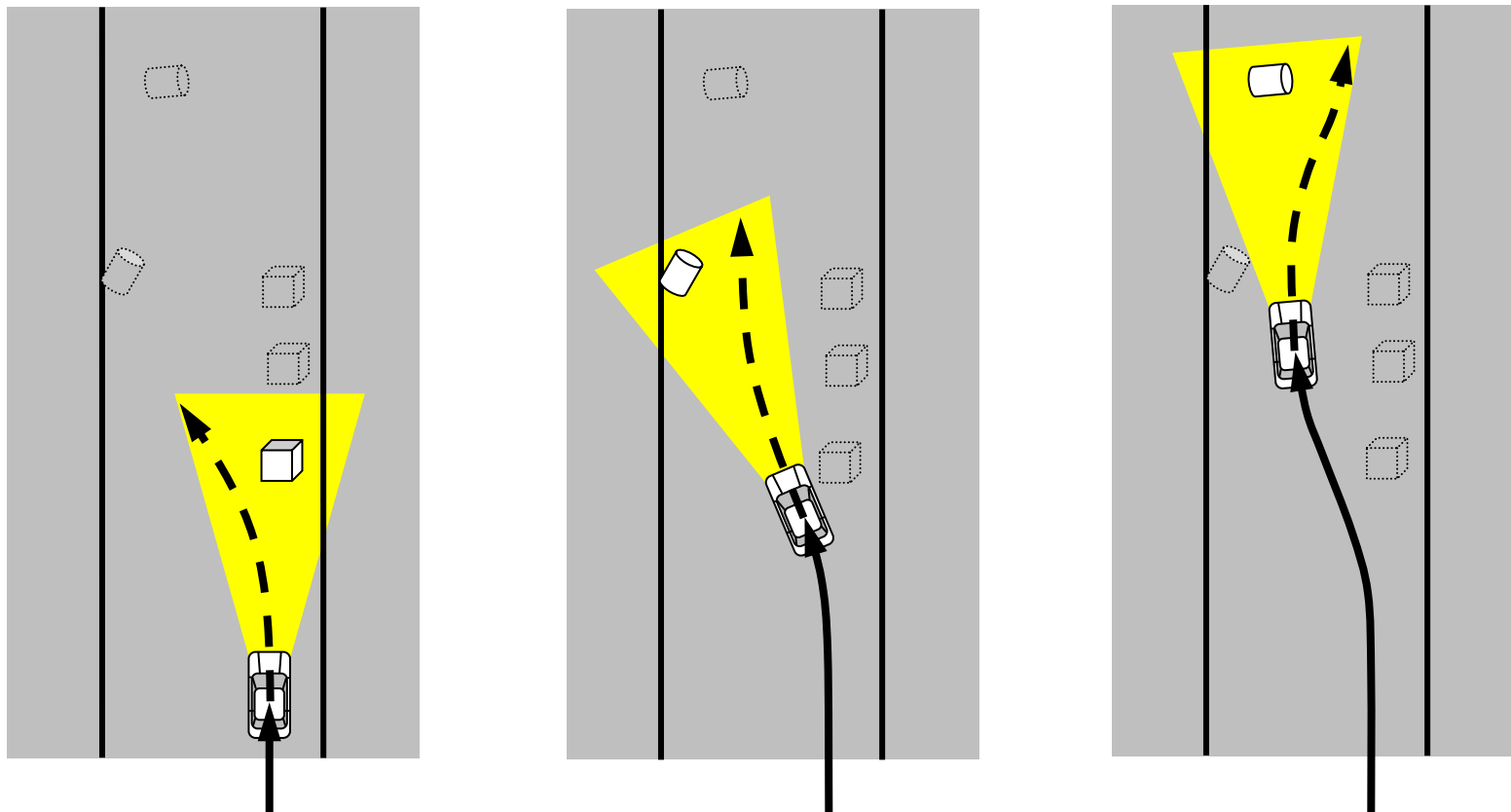
Feedback control by **real-time optimization** of the system response over a finite future.



Control action is updated **at each time** to utilize new information.

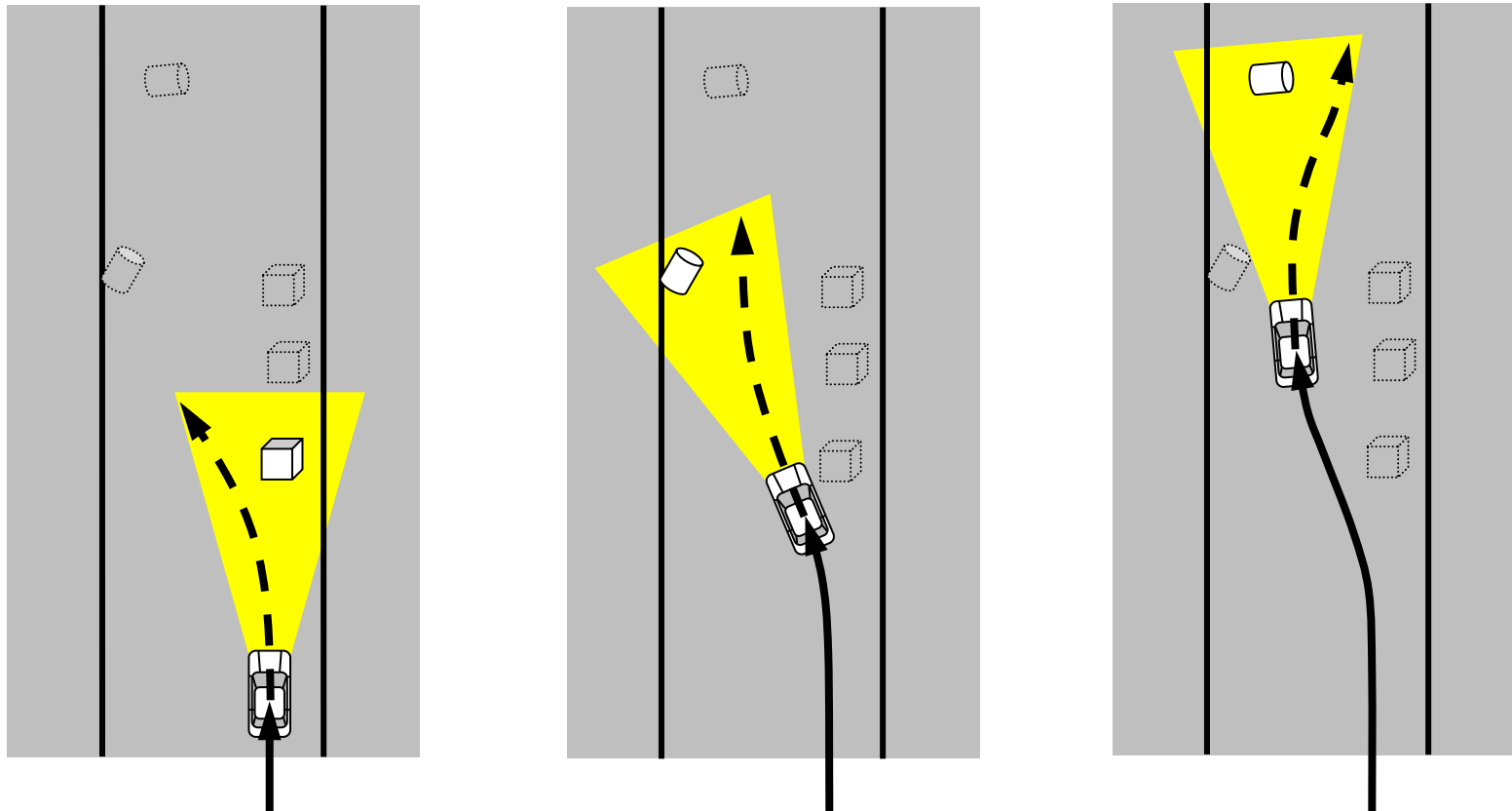
What is MPC?

Feedback control by **real-time optimization** of the system response over a finite future.

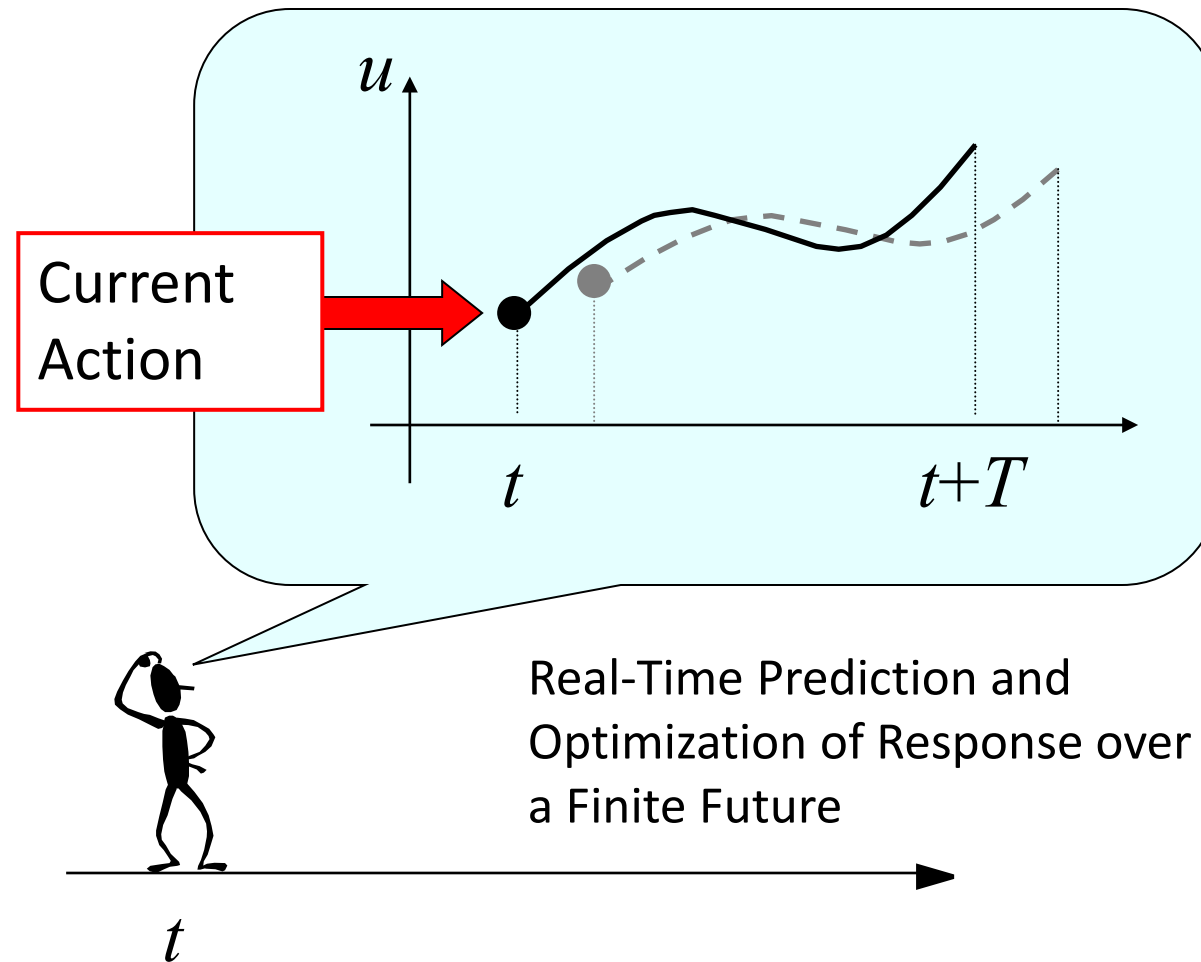


What is MPC?

A General Framework for Feedback Control of Nonlinear Dynamical Systems



Receding Horizon



OCP in MPC (Nonlinear MPC; NMPC)

State Equation and Constraint

$$C \leq 0 \Leftrightarrow C + v^2 = 0$$

$$\dot{x}(t) = f(x(t), u(t), t), \quad C(x(t), u(t), t) = 0$$

Performance Index

Receding Horizon

$$J = \varphi(x(t+T), t+T) + \int_t^{t+T} L(x(\tau), u(\tau), \tau) d\tau$$

NMPC as **State Feedback**

Initial Value of Optimal Control
over $t \leq \tau \leq t+T$

$$u(t) = u_{opt}(t; x(t), t)$$

Parameterized OCP in MPC

Fixed Horizon over $0 \leq \tau \leq T$

$$\frac{dx^*(\tau; t)}{d\tau} = f(x^*(\tau; t), u^*(\tau; t), \tau + t), \quad x^*(0; t) = x(t)$$

$$C(x^*(\tau; t), u^*(\tau; t), \tau + t) = 0$$

Actual State at t

$$J = \varphi(x^*(T; t), T + t) + \int_0^T L(x^*(\tau; t), u^*(\tau; t), \tau + t) d\tau$$

Actual Input to the System

$$u(t) = u_{opt}^*(0; t)$$

Depends on initial condition
 $x^*(0; t) = x(t)$ (actual state)

History of MPC

- Early work (Coales&Noton '56), (Propoi '63), (Merriam '64)
- LQR variant (Kleinmann '70), (Thomas '75), (Kwon&Pearson '77)
- Process Control:
PFC (Richalet et al. '78), DMC (Cutler&Ramaker '80), QDMC (Garcia&Morshedi '86),
GPC (Clarke et al. '87)
- Stability of NMPC (Chen&Shaw '82), (Mayne&Michalska '90), (Chen&Allgöwer '98),
(Jadbabaie et al. '01), etc.
- Numerical methods (90s-): mp-QP, RTO, etc.

Parameterized ELE in MPC

$$\frac{dx^*(\tau;t)}{d\tau} = f(x^*(\tau;t), u^*(\tau;t), \tau + t), \quad x^*(0;t) = x(t)$$

$$\frac{d\lambda^*(\tau;t)}{d\tau} = -\nabla_x H(x^*(\tau;t), u^*(\tau;t), \lambda^*(\tau;t), \tau + t),$$

$$\lambda^*(T;t) = \nabla_x \varphi(x^*(T;t), T + t)$$

$$\nabla_u H(x^*(\tau;t), u^*(\tau;t), \lambda^*(\tau;t), \tau + t) = 0$$

Actual State at t

Nonlinear TPBVP

Actual Input to the System

$$u(t) = u_{opt}^*(0;t)$$

HJBE in MPC

Value Function

$$V(x, \tau; t) = \min_{u[\tau, T]} \left[\varphi(x^*(T; t), T + t) + \int_{\tau}^T L(x^*(\tau'; t), u^*(\tau'; t), \tau' + t) d\tau' \right]$$

HJBE

$$-\nabla_{\tau} V(x, \tau; t) = \min_u H(x, u, \nabla_x V(x, \tau; t), \tau + t), \quad V(x, T; t) = \varphi(x, T + t)$$

Optimal Control

Nonlinear PDE

$$u_{opt}^*(x, \tau; t) = \arg \min_u H(x, u, \nabla_x V(x, \tau; t), \tau + t),$$
$$u_{MPC}(x, t) = u_{opt}^*(x, 0; t)$$

In a time-invariant case (time t does not appear explicitly):

$$u_{MPC}(x) = u_{opt}^*(x, 0)$$

Difficulty in MPC: Solution Methods

ELE

- Open-loop solution for a given state
- Iterative methods: gradient methods, Newton's method (computationally demanding for real-time implementation)

HJBE

- Closed-loop solution (state feedback)
- Explosive growth of computation and storage for high dimensional systems (curse of dimensionality)

Implementation of MPC for **fast and complex nonlinear systems** with sampling periods of the order of **milliseconds** is challenging!

Difficulty in MPC: Closed-Loop Stability

Closed-loop stability of (time-invariant) MPC with a finite horizon?

Key Idea: Value function $V_{MPC}(x) = V(x, 0)$ can be a Lyapunov function.

From HJBE

$$\dot{V}_{MPC}(x) = -L(x, u_{MPC}(x)) - \nabla_{\tau} V(x, 0)$$

$$u_{MPC}(x) = u_{opt}(x, 0)$$

Methods for Guaranteeing Stability

- Terminal constraint $x^*(T; t) = 0$
- Terminal constraint $x^*(T; t) \in \Omega$, a stabilizing feedback to make Ω invariant, and a bound on the infinite horizon cost in Ω
- Control Lyapunov function (CLF) as the terminal penalty $\varphi(x)$ and a feedback $u = k(x)$ such that

$$\frac{\partial \varphi(x)}{\partial x} f(x, k(x)) \leq -L(x, k(x))$$

Stability guarantee of MPC is challenging!

Some Variants of MPC

- Robust MPC (Constraint satisfaction under uncertainty)
- Stochastic MPC (Open-loop/closed-loop optimization, chance constraint)
- Adaptive MPC
- Learning MPC (Kabzan et al. '19; Gros&Zanon '20; Rosolia&Borrelli '20)
Cf. MPC in Learning: MPC-guided policy search (Zhang et al. '16)
- Distributed MPC
- Output MPC
- Moving Horizon Estimation (Fitting model and measurement over a finite past)

J. B. Rawlings, et al.: Model Predictive Control: Theory, Computation, and Design, 2nd Ed., Nob Hill Publishing (2017)

A. Mesbah: Stochastic Model Predictive Control: An Overview and Perspectives for Future Research; IEEE Control Systems Magazine, 36(6), 34-44 (2016)

Summary

- MPC: State feedback control by real-time optimization over a finite future
- General framework for feedback control of nonlinear systems
- Difficulties: Solution methods and stability guarantee
- Implementation of MPC for **fast and complex nonlinear systems** with sampling periods of the order of **milliseconds** is challenging!
- There are some variants of MPC.

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NeurIPS2021 Tutorial

Real-Time Optimization for Fast and Complex Control Systems

Part 3

Real-Time Optimization for Model Predictive Control



Toshiyuki Ohtsuka

Department of Systems Science

Graduate School of Informatics

Kyoto University

Outline

Part 1: Introduction to Control Systems

Part 2: Optimal Control and Model Predictive Control

Part 3: Real-Time Optimization for Model Predictive Control

Part 4: Advanced Topics in Model Predictive Control

Outline of Part 3

- Real-Time Optimization (RTO) Algorithm for Nonlinear MPC (NMPC)
- Automatic Code Generation Tool

Solution Methods of MPC

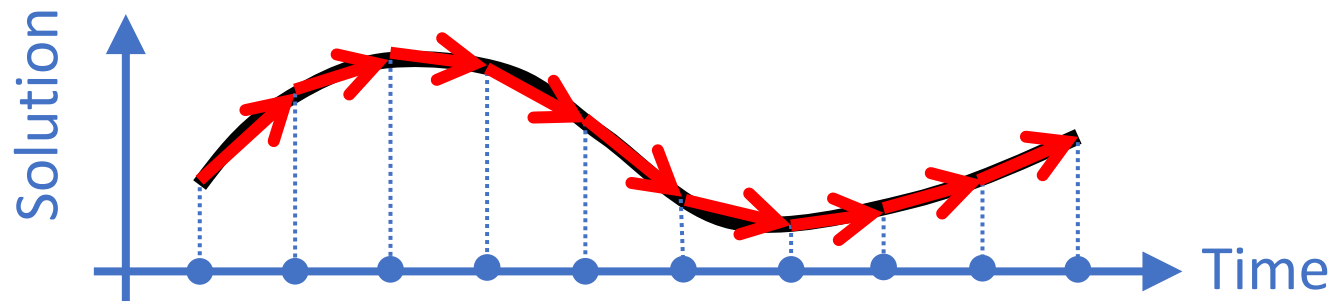
- **Off-Line Methods (Explicit MPC)**
 - (Bemporad et al. '02), (Zeilinger et al. '11), etc.
- **On-Line Methods (Real-Time Optimization; RTO)**
 - (O '97), (O '04) **Real-Time Algorithm as ODE**
 - (Diehl et al. '05), (Alamir '06), (DeHaan&Guay '07), (Zavala&Biegler '09), (Graichen '12), (Stella et al. '17), etc.
 - (Deng&O '19) **Parallel Algorithm**

Numerical Algorithm for NMPC

An **iterative search** of an optimal solution within a **short sampling period** may fail to converge.

However...

The **small change** in the optimal solution during the short sampling period can be traced **without an iterative search!**



Numerical Algorithm for NMPC

An **iterative search** of an optimal solution within a **short sampling period** may fail to converge.

However...

The **small change** in the optimal solution during the short sampling period can be traced **without an iterative search!**

Only one linear equation at each sampling time: Continuation/GMRES

(O '04)

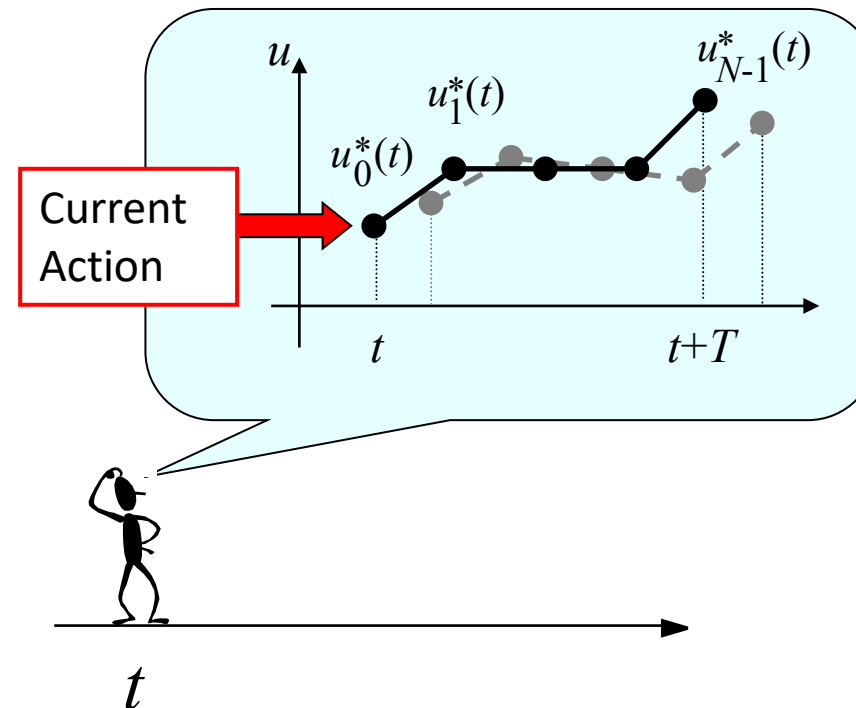
Discretization of the Horizon

Input Sequence over the Horizon at Time t :

$$U(t) = [u_0^{*T}(t) \quad \mu_0^{*T}(t) \quad \cdots \quad u_{N-1}^{*T}(t) \quad \mu_{N-1}^{*T}(t)]^T$$

Actual Input: $u(t) = u_0^*(t)$, Lagrange Multiplier: $\mu_i^*(t)$

Discretization Step $\Delta\tau = \frac{T}{N}$
 \neq Sampling Period Δt



Discretized ELE

$$\begin{aligned}x_{i+1}^*(t) &= x_i^*(t) + f(x_i^*(t), u_i^*(t))\Delta\tau, & x_0^*(t) &= x(t) \\ \lambda_i^*(t) &= \lambda_{i+1}^*(t) + \nabla_x H(x_i^*(t), u_i^*(t), \lambda_{i+1}^*(t), \mu_i^*(t))\Delta\tau \\ & & \lambda_N^*(t) &= \nabla_x \varphi(x_N^*(t)) \\ \nabla_u H(x_i^*(t), u_i^*(t), \lambda_{i+1}^*(t), \mu_i^*(t)) &= 0 \\ C(x_i^*(t), u_i^*(t)) &= 0\end{aligned}$$

Hamiltonian: $H(x^*, u^*, \lambda^*, \mu^*) = L(x^*, u^*) + \lambda^{*T} f(x^*, u^*) + \mu^{*T} C(x^*, u^*)$

State and costate are functions of $U(t)$ and $x_0^*(t) = x(t)$

Nonlinear Equation for the Input Sequence

Future Input Sequence

Current State

$$F(U(t), x(t), t) = \begin{bmatrix} \nabla_u H(x_0^*(t), u_0^*(t), \lambda_1^*(t), \mu_0^*(t)) \\ C(x_0^*(t), u_0^*(t)) \\ \vdots \\ \nabla_u H(x_{N-1}^*(t), u_{N-1}^*(t), \lambda_N^*(t), \mu_{N-1}^*(t)) \\ C(x_{N-1}^*(t), u_{N-1}^*(t)) \end{bmatrix} = 0$$

State and costate are functions of $U(t)$ and $x_0^*(t) = x(t)$

Continuation Method

Condition for Stabilizing $F(U(t), x(t), t) = 0$

$$\frac{d}{dt}F(U(t), x(t), t) = -\zeta F(U(t), x(t), t) \quad (\zeta > 0)$$



Linear Equation for $\dot{U}(t)$:

$$\frac{\partial F}{\partial U} \dot{U} = -\zeta F - \frac{\partial F}{\partial x} \dot{x} - \frac{\partial F}{\partial t}$$



Update of Optimal Solution $U(t)$:

$$U(t + \Delta t) = U(t) + \dot{U}(t)\Delta t$$

No Iterative Search

Continuatin/GMRES Method

1) At each time t , measure the state $x(t)$

2) Solve a linear equation for $\dot{U}(t)$:

$$\frac{\partial F}{\partial U} \dot{U} = -\zeta F - \frac{\partial F}{\partial x} \dot{x} - \frac{\partial F}{\partial t}$$

3) Update $U(t)$ by $U(t + \Delta t) = U(t) + \dot{U}(t)\Delta t$

4) Go to Step 1)

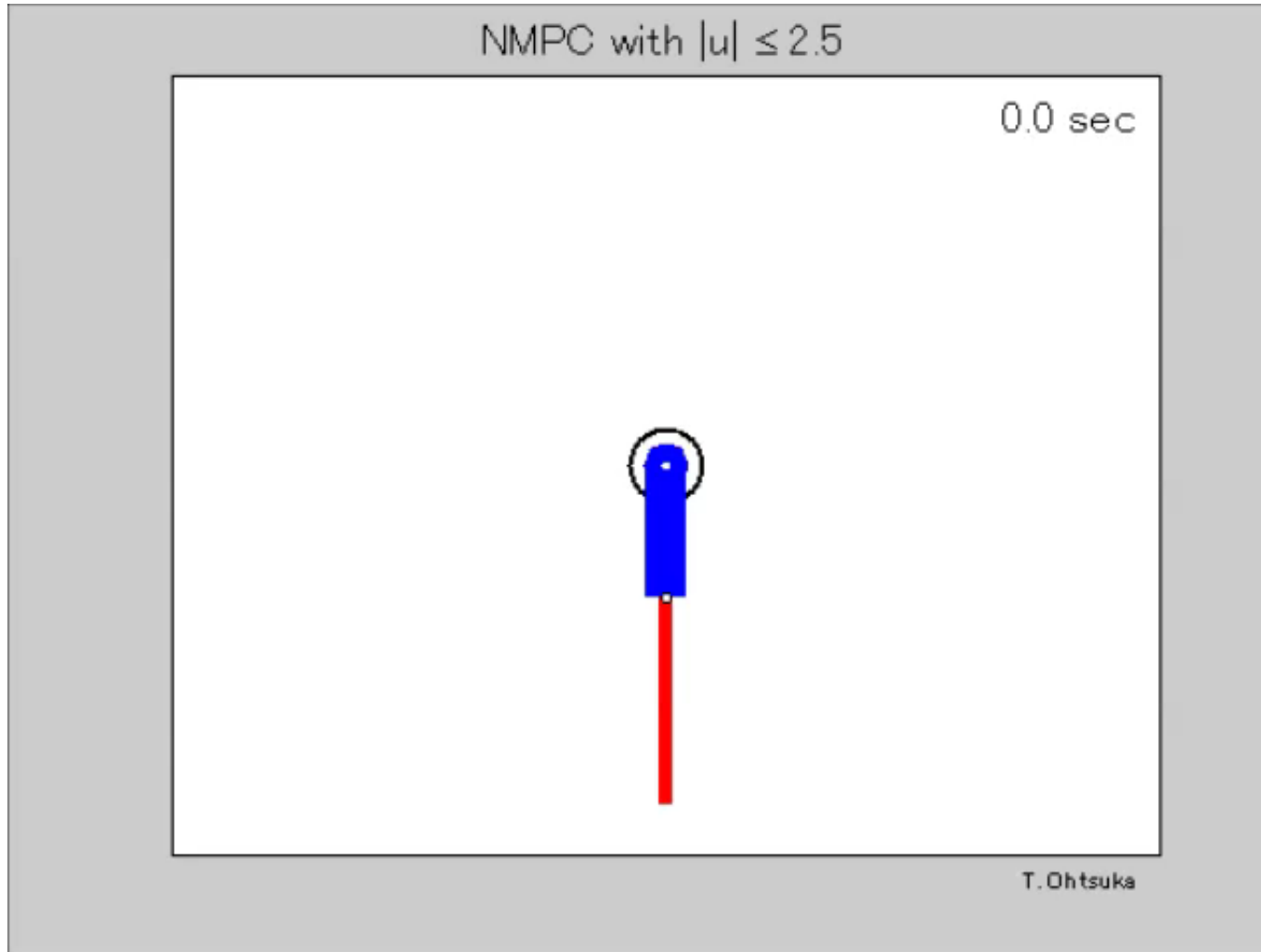
- Only one linear equation at each time
- Efficient linear equation solver **GMRES**
- Horizon length: $T(0) = 0, T(t) \rightarrow T_f$ (easy to find $U(0)$)

Jacobian-Free

$\frac{\partial F}{\partial U}$ unnecessary

(O'04)

Two-Link Arm



- Free Elbow Joint and Constrained Shoulder Torque
- Computational Time per Update: 0.5ms (CPU: Core 2 Duo 1.2GHz)
(O'12)

Underactuated Hovercraft



- Fixed On-Off Thrusters, No Lateral Thrust
- Sampling Period $1/120$ s (CPU: Athlon 900 MHz)

(Seguchi&O '03)

Automatic Ship Maneuvering

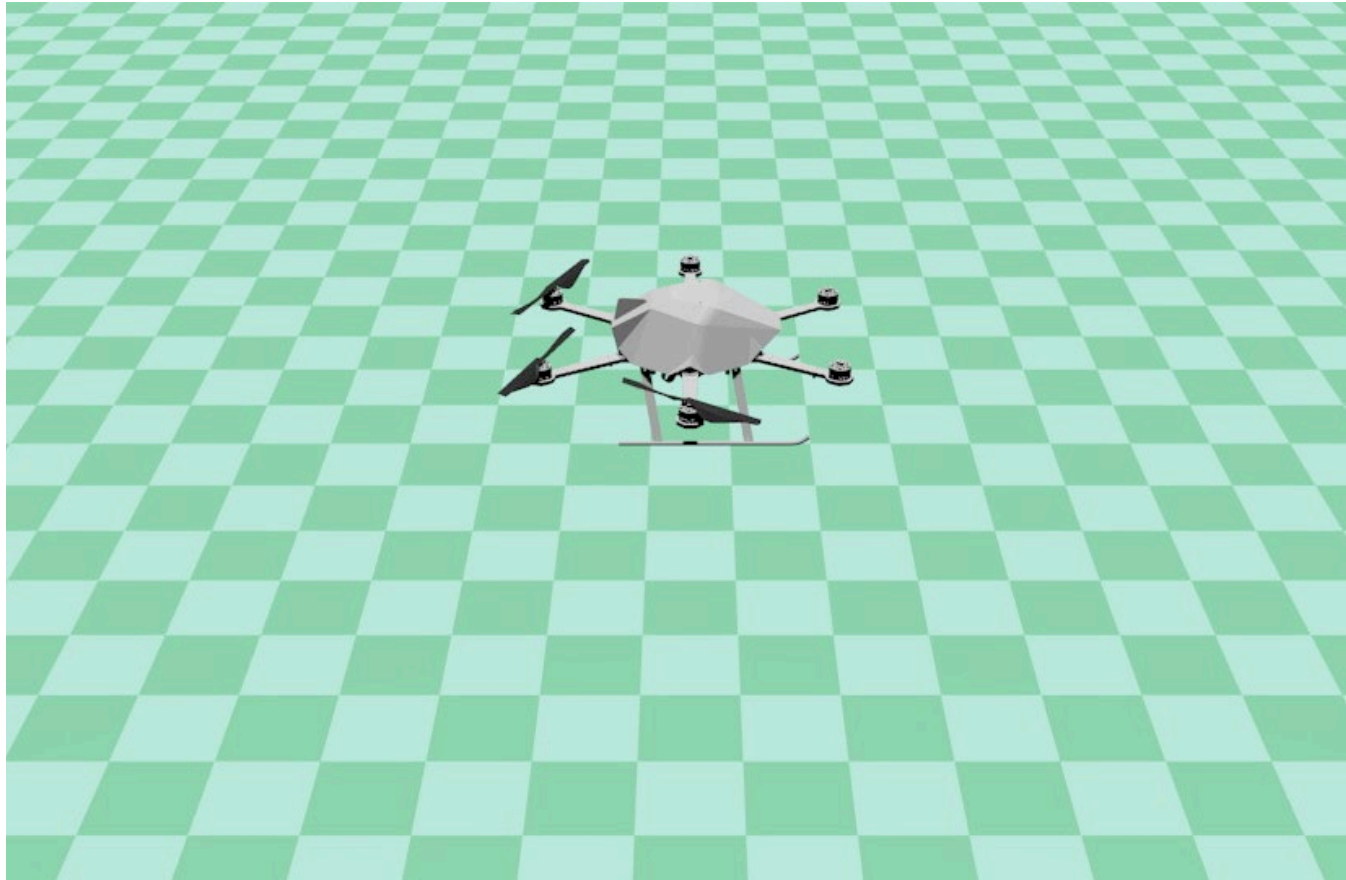


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- Redundant Actuators
- NMPC for Route Tracking and Thrust Assignment

(Hamamatsu et al. '08)

Hexacopter with Failed Rotors



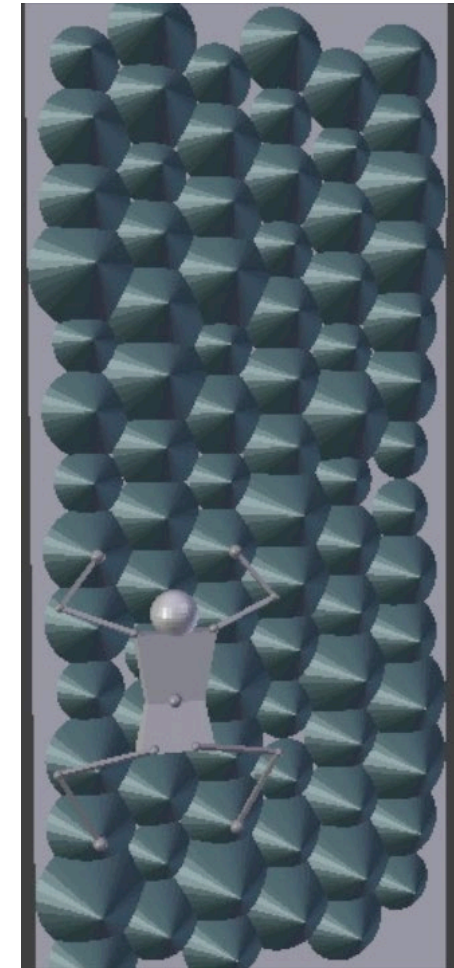
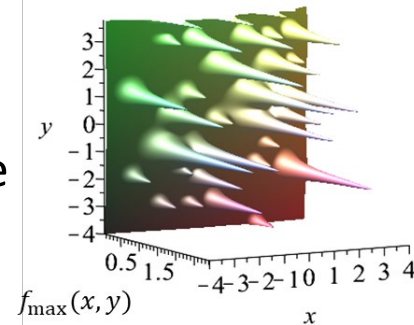
- Nonlinear Model with Quaternions
- Constraints on Thrusts
- Same NMPC for All Failure Patterns with Different Weights
- Computational Time per Update: below 0.2ms

(Aoki et al. '21)

Climbing Humanoid Robot

- Integrated optimization of path and motions
- Penalty on deviation from reference velocity
- Constraints on force/moment balance
- Constraints on holding forces
- 12 states, 36 inputs (including dummy inputs), and 15 constraints
- Computation time per update < 20 ms (Core i5, 1.8GHz)

Maximum
Holding Force
on the Wall



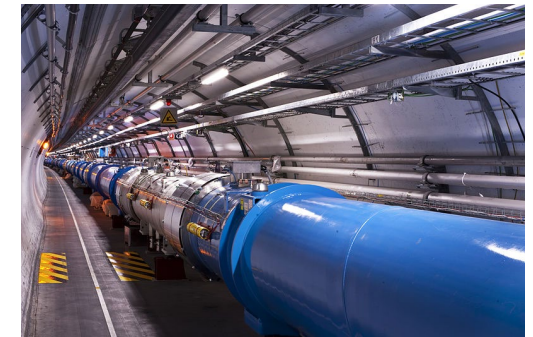
(Omoto et al. '19)

Other Applications

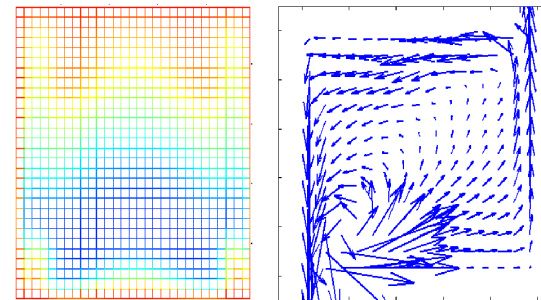
- Tethered Satellite, Robots, Automobiles, Formation Flight of Helicopters, FOWT, etc.
- Flight Experiment of On-Line Path Generation for an Aircraft
- Temperature Control for Superconducting Magnets in the Large Hadron Collider (LHC)
- Distributed Parameter Systems (Thermofluid Systems)



© JAXA



Maximilien Brice (CERN)
(https://commons.wikimedia.org/wiki/File:Views_of_the_LHC_tunnel_sector_3-4,_tirage_2.jpg), „Views of the LHC tunnel sector 3-4, tirage 2“, <https://creativecommons.org/licenses/by-sa/3.0/legalcode>



Other Problems

- Nonlinear Receding Horizon Differential Game
 - Minimizer (Control) and Maximizer (Disturbance) (Hirota et al. '17)
 - Nash equilibrium of Multiple Players (Azuma&O '11)
- Moving Horizon Estimation (Soneda&O '05)
 - Optimal Fitting of a Model and Measurements over a Finite Past

Outline of Part 3

- Real-Time Optimization (RTO) Algorithm for Nonlinear MPC (NMPC)
- Automatic Code Generation Tool

Software Tools for NMPC

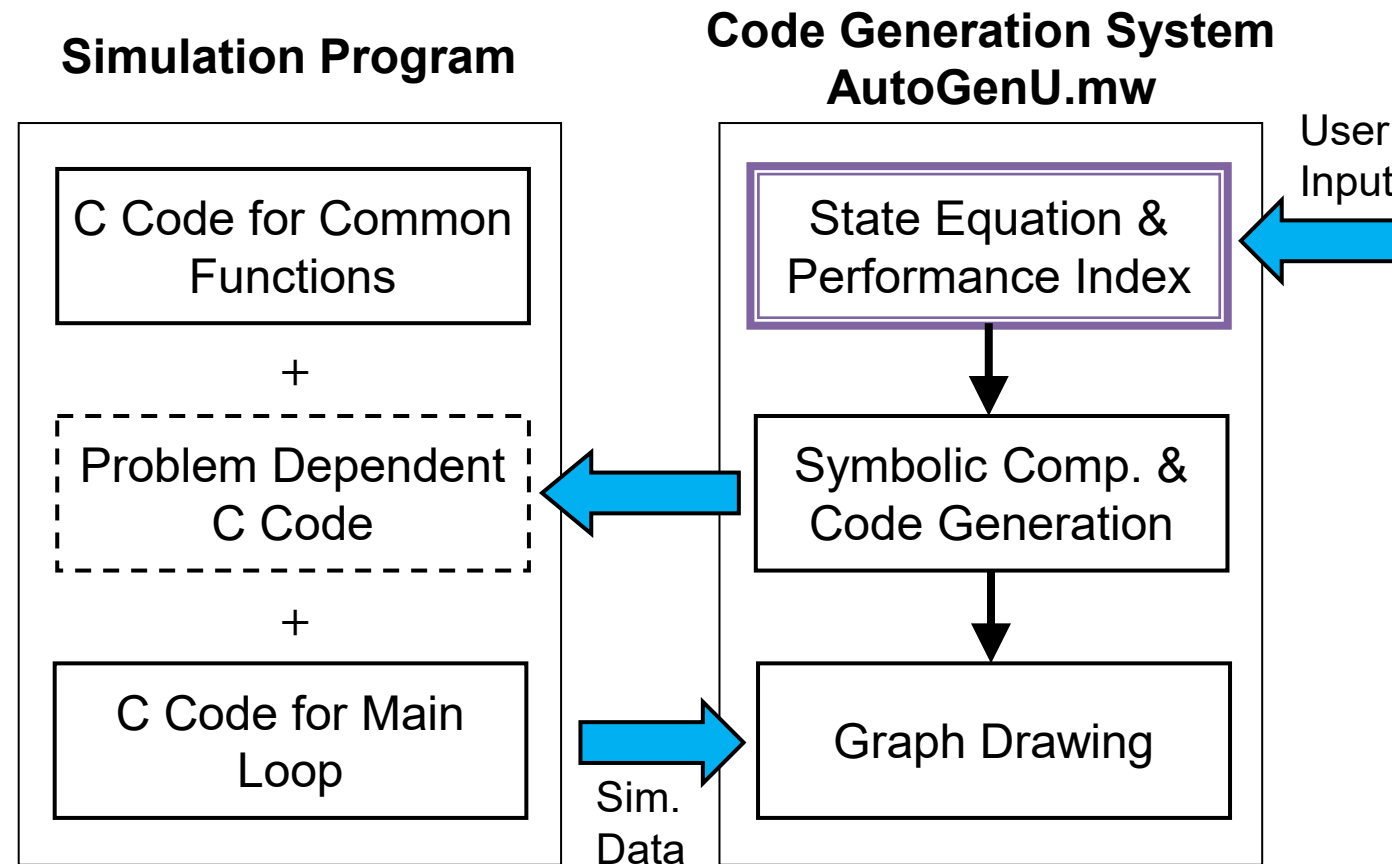
- **MPT** (Kvasnica et al. '04), **MPT3** (Herceg et al. '13)
- **AutoGen** (O&Kodama '02)
- **AutoGenU** for Mathematica (O '04), **Maple** (O '15), **Python Jupyter Notebook** (Katayama&O '20)
- **ACADO** (Houska et al. '11), **acados** (Verschueren et al., '21)
- **CVXGEN** (Mattingley&Boyd '12), **qpOASES** (Ferreau et al. '14)
- **FORCES PRO**
- **GRAMPC** (Englert et al. '19)
- **ParNMPC** (Deng&O '20)

AutoGenU for Maple

- Developed by Cybernet Systems in collaboration with the speaker
- Symbolic computation by Maple
- C code generation
- Integrated compilation, execution, and graph plotting
- Freely available at Maplesoft Application Center website

Automatic Code Generation

C code for simulation is generated from the state equation and performance index specified in **Maple**.



Example: Semi-Active Damper

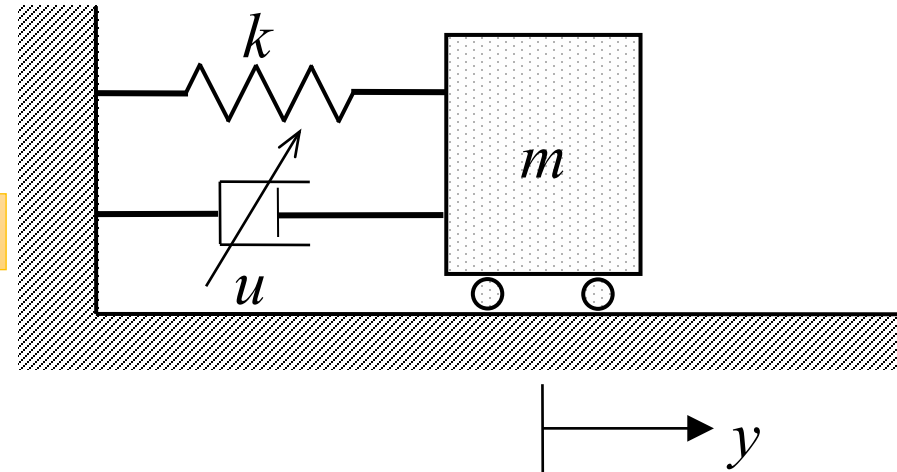
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} x_2 \\ ax_1 + bx_2u \end{bmatrix}$$

Bilinear

$$0 \leq u \leq u_{max}$$

$$\left(u_1 - \frac{u_{max}}{2}\right)^2 + u_2^2 - \frac{u_{max}^2}{4} = 0$$

$$J = \frac{1}{2}x^T(t+T)S_f x(t+T) + \int_t^{t+T} \left(\frac{1}{2}x^T(\tau)Qx(\tau) + \frac{r_1}{2}u_1^2(\tau) - r_2u_2 \right) d\tau$$



$x_1 = y, x_2 = \dot{y},$
 $u_1 = u, u_2: \text{Dummy Input}$

Maple Worksheet AutoGenU.mw

AutoGenU for Maple
(GUI version)

- ▶ Introduction
- ▶ Initialize
- ▶ Define Setting Parameters
- ▶ Function for C Code Generation
- ▶ Generate Euler-Lagrange Equations
- ▶ Generate C Code
- ▶ Run Simulation
- ▶ Show Graphs

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ID Number: 201309191000

Ready Editable Maple デフォルトファイル G:\ohtsuka\AutoGen\AutoGenU\RT0\MapleVersionDemo Memory: 418M Time: 0.04s Zoom: 100% Text Mode

State Equation and Constraint

Input	
<code>fxu</code>	<code><x[2], a * x[1] + b * x[2] * u[1]></code>
<code>Cxu</code>	<code><(u[1] - umax/2)^2 + u[2]^2 - (umax/2)^2></code>

Input Box

Result	
	$fxu := \begin{bmatrix} x_2 \\ b u_1 x_2 + a x_1 \end{bmatrix}$
	$Cxu := \begin{bmatrix} \left(u_1 - \frac{1}{2} umax\right)^2 + u_2^2 - \frac{1}{4} umax^2 \end{bmatrix}$

Performance Index

Input

L	$(xv^T Q xv) / 2 + r[1] * u[1]^2 / 2 - r[2] * u[2]$
----------	---

phi	$(xv^T S f xv) / 2$
------------	---------------------

Result

$$L := \frac{1}{2} x_1^2 q_1 + \frac{1}{2} x_2^2 q_2 + \frac{1}{2} r_1 u_1^2 - r_2 u_2$$
$$\phi := \frac{1}{2} x_1^2 s f_1 + \frac{1}{2} x_2^2 s f_2$$

User's Variables and Arrays

Input	
MyVarNames	<code>["a", "b", "umax"]</code>
MyVarValues	<code>[-1, -1, 1]</code>
MyArrNames	<code>["q", "r", "sf"]</code>
MyArrDims	<code>[dimx, dimu, dimx]</code>
MyArrValues	<code>[<1,10>, <1,0.01>, <1,10>]</code>

Variables

Arrays

Simulation Conditions

tsim	<input type="text" value="20.0"/>	Final Time of Simulation
ht	<input type="text" value="0.001"/>	Time Step for Simulation
tf	<input type="text" value="1.0"/>	Horizon Length
x0	<input type="text" value="<2, 0>"/>	Initial State
kmax	<input type="text" value="5"/>	No. of Iterations in GMRES
dv	<input type="text" value="50"/>	No. of Time Steps in Horizon, N

Maple Worksheet AutoGenU.mw

Maple 2020 interface showing the worksheet 'AutoGenU.mw'. The main content area displays a table of contents with the following items:

- ▶ Introduction
- ▶ Initialize
- ▶ Define Setting Param
- ▶ Function for C Code Generation
- ▶ Generate Euler-Lagrange Equations
- ▶ Generate C Code
- ▶ Run Simulation
- ▶ Show Graphs

A yellow callout box highlights the following items:

- ▶ Generate Euler-Lagrange Equations
- ▶ Generate C Code

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Symbolic Computation for ELE

```
> H:=L + lmdv^%T.fxu + muv^%T.Cxu;
```

$$H := \frac{1}{2} x_1^2 q_1 + \frac{1}{2} x_2^2 q_2 + \frac{1}{2} r_1 u_1^2 - r_2 u_2 + lmd_1 x_2 + lmd_2 (b u_1 x_2 + a x_1) + u_3 \left(\left(u_1 - \frac{1}{2} umax \right)^2 + u_2^2 - \frac{1}{4} umax^2 \right)$$

Hamiltonian

```
> uv := convert([uv, muv], Vector);
```

$$uv := \begin{bmatrix} u_1 \\ u_2 \\ u_3 \end{bmatrix}$$

```
> dimuc := dimu + dimc;
```

```
> Hx := Dv(H, xv);
```

$$Hx := \begin{bmatrix} a lmd_2 + q_1 x_1 \\ b lmd_2 u_1 + q_2 x_2 + lmd_1 \end{bmatrix}$$

```
> Hu := Dv(H, uv);
```

$$Hu := \begin{bmatrix} r_1 u_1 + lmd_2 b x_2 + u_3 (2 u_1 - umax) \\ 2 u_2 u_3 - r_2 \\ \left(u_1 - \frac{1}{2} umax \right)^2 + u_2^2 - \frac{1}{4} umax^2 \end{bmatrix}$$

Partial Derivatives

Maple Worksheet AutoGenU.mw

AutoGenU for Maple
(GUI version)

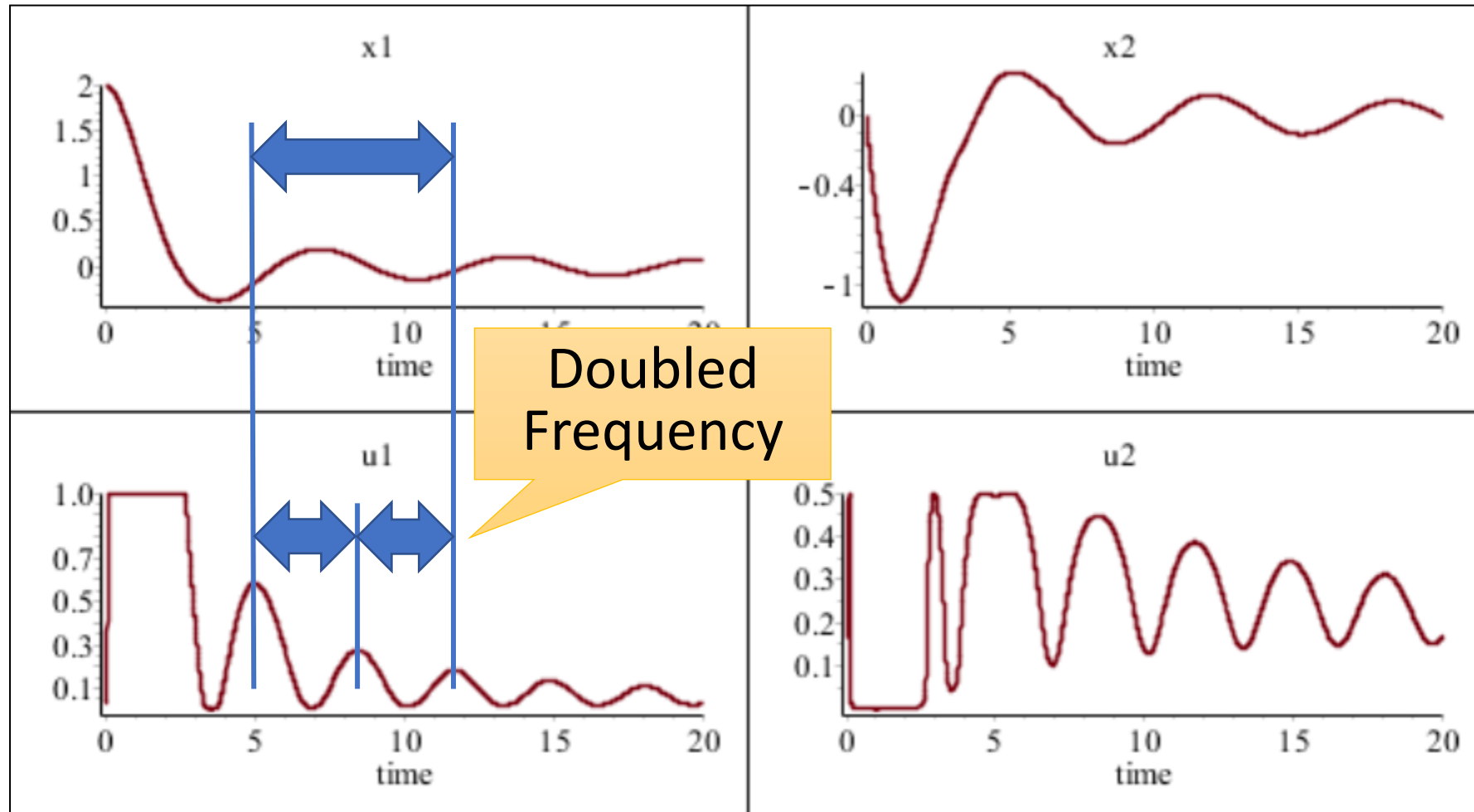
- ▶ Introduction
- ▶ Initialize
- ▶ Define Setting Parameters
- ▶ Function for C Code Generation
- ▶ Generate Euler-Lagrange Equations
- ▶ Generate C Code
- ▶ Run Simulation
- ▶ Show Graphs

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Run Simulation

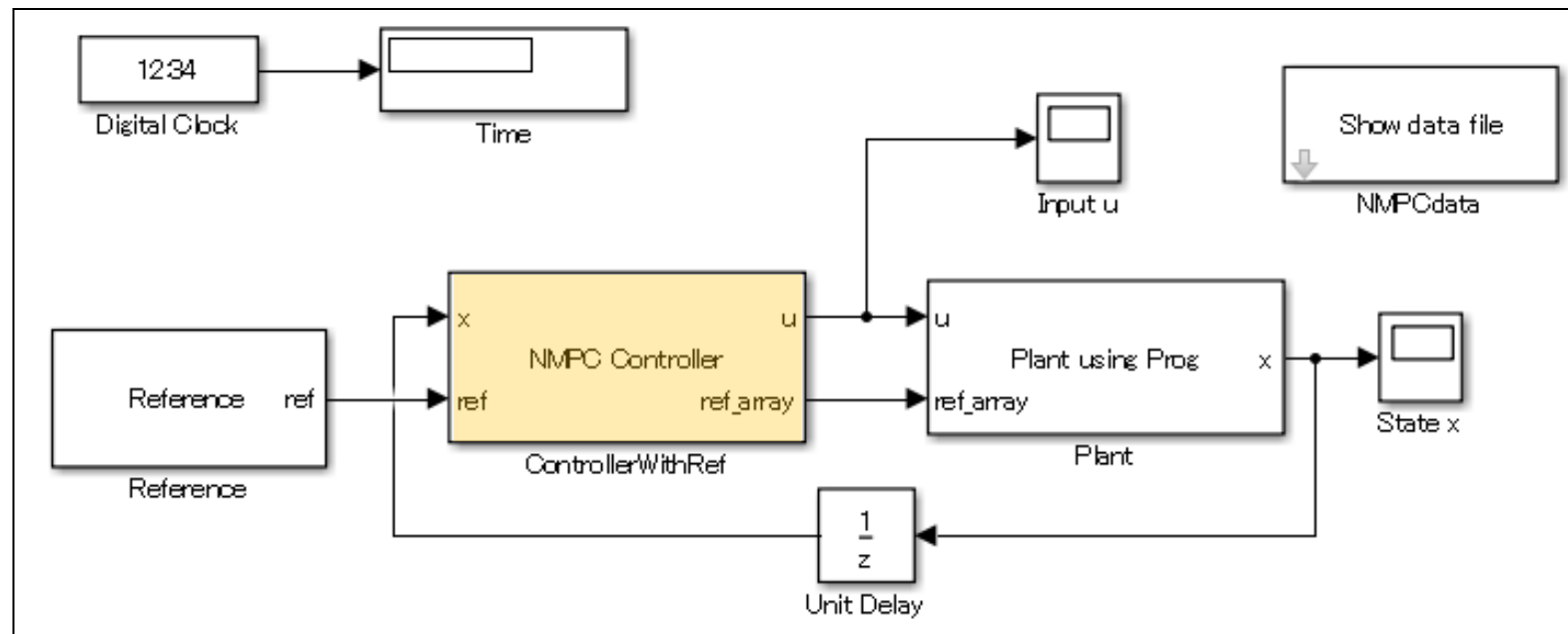
Show Graphs

Simulation Result



MPC Design & Connector

- Automatic Generation of a **Simulink Block**
- Contact Cybernet Systems for license.
Email: infomaple@cybernet.co.jp



Summary

- **Nonlinear model predictive control (NMPC)** is a very general framework for feedback control of nonlinear systems.
- **Real-time optimization (RTO)** is a key component of NMPC.
- **C/GMRES** is a continuation-based real-time algorithm for NMPC with successful applications.
- **AutoGenU for Maple** is a Maple worksheet for automatic C code generation of C/GMRES.
- RTO algorithms can be applied not only to MPC but also to various problems such as estimation, differential games, etc.

URLs

- Maplesoft Application Center:
<http://www.maplesoft.com/applications/view.aspx?SID=153555>
- Maplesoft Webinar:
<https://www.youtube.com/watch?v=fVsYNjQfYUg>
- MPC Design & Connector License: infomaple@cybernet.co.jp
- Mathematica Version: <http://www.ids.sys.i.kyoto-u.ac.jp/~ohtsuka/code/autogenu/autogenu.zip>
- Python Jupyter Notebook: <https://github.com/mayataka/CGMRES>

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NeurIPS2021 Tutorial

Real-Time Optimization for Fast and Complex Control Systems

Part 4

Advanced Topics in Model Predictive Control



Toshiyuki Ohtsuka

Department of Systems Science

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Kyoto University

Outline

Part 1: Introduction to Control Systems

Part 2: Optimal Control and Model Predictive Control

Part 3: Real-Time Optimization for Model Predictive Control

Part 4: Advanced Topics in Model Predictive Control

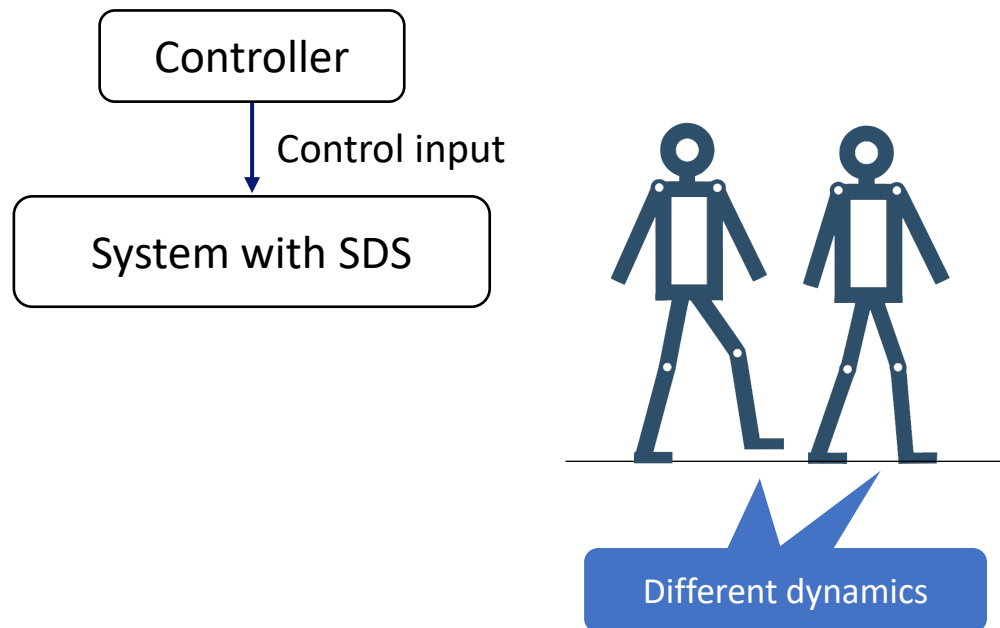
Outline of Part 4

- NMPC with State-Dependent Switches and State Jumps
- Parallel Algorithm for NMPC

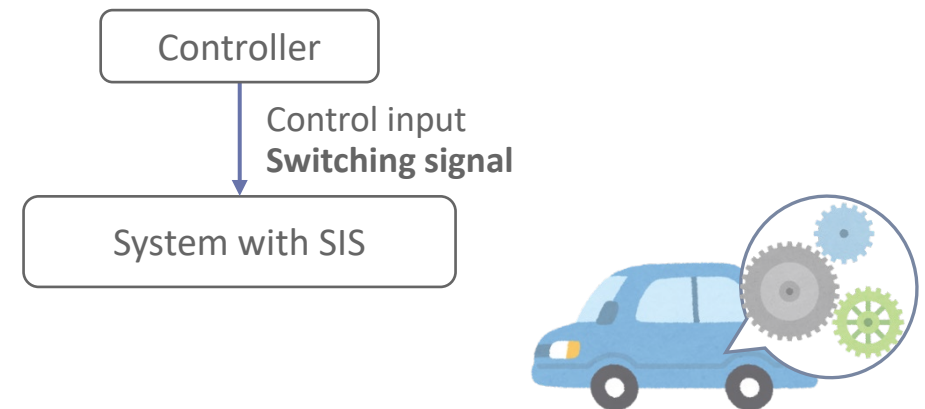
S. Katayama, M. Doi, T. Ohtsuka: A Moving Switching Sequence Approach for Nonlinear Model Predictive Control of Switched Systems with State-Dependent Switches and State Jumps; *Int. J. Robust & Nonlinear Contr.*, 30(2), 719-740 (2020)
H. Deng, T. Ohtsuka: A Parallel Newton-type Method for Nonlinear Model Predictive Control; *Automatica*, 109, paper 108560 (2019)

State-Dependent Switches of Dynamics

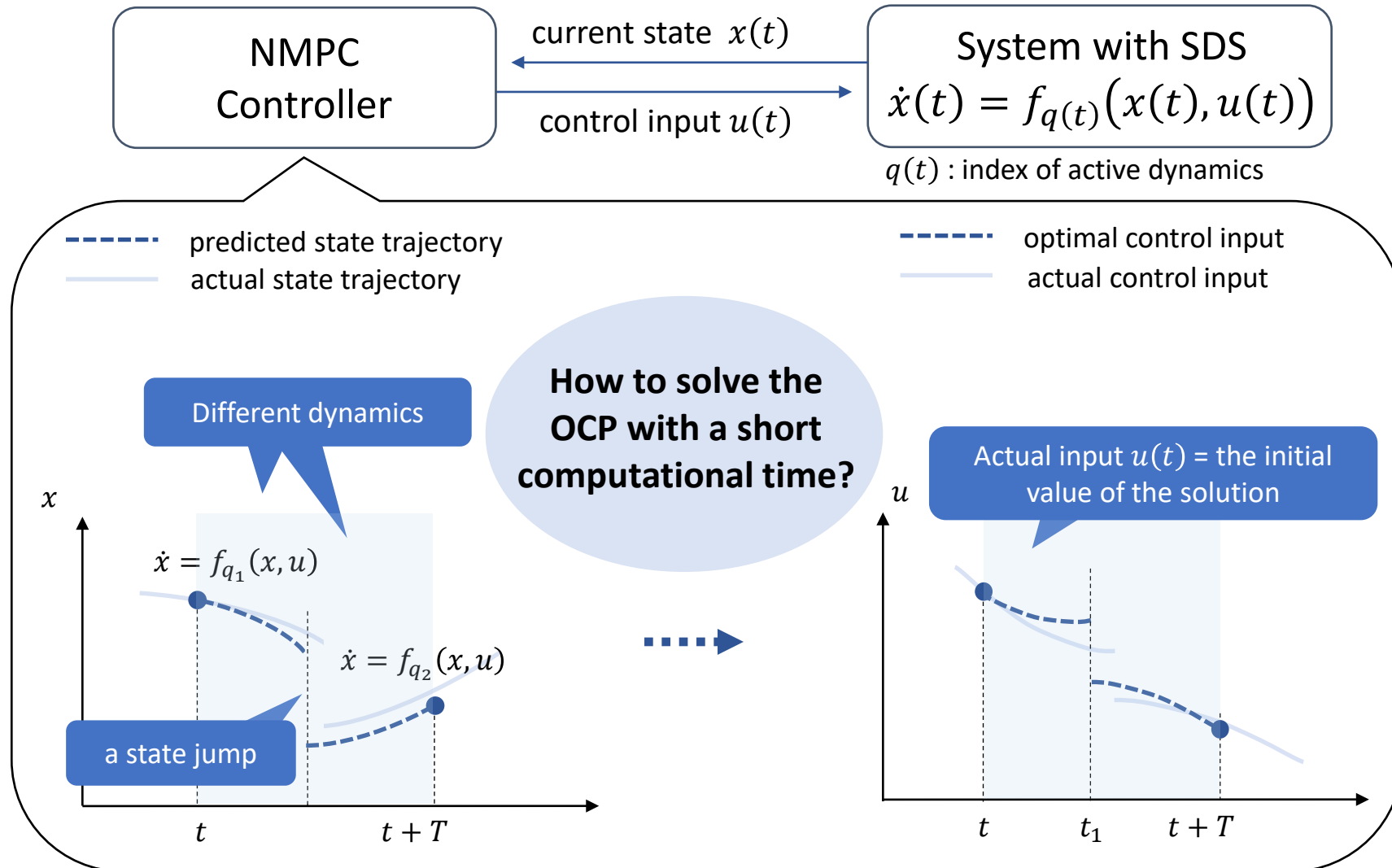
- Systems with state-dependent switches (SDS)



Cf. Systems with state-independent switches (SIS)

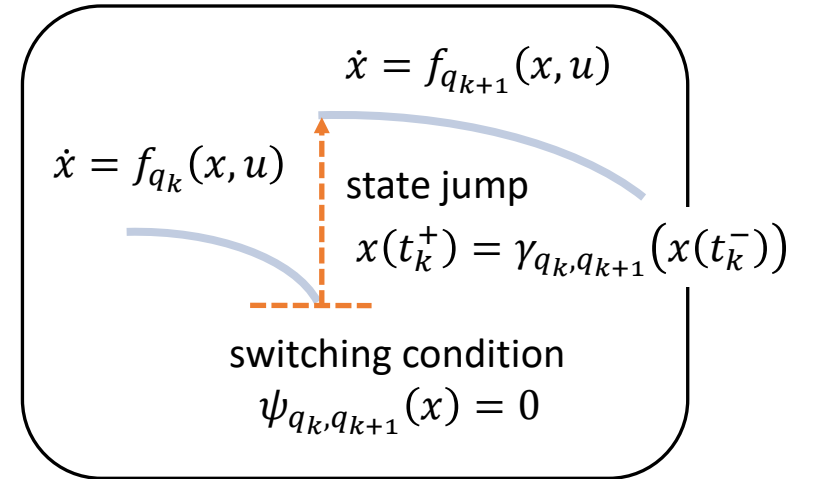
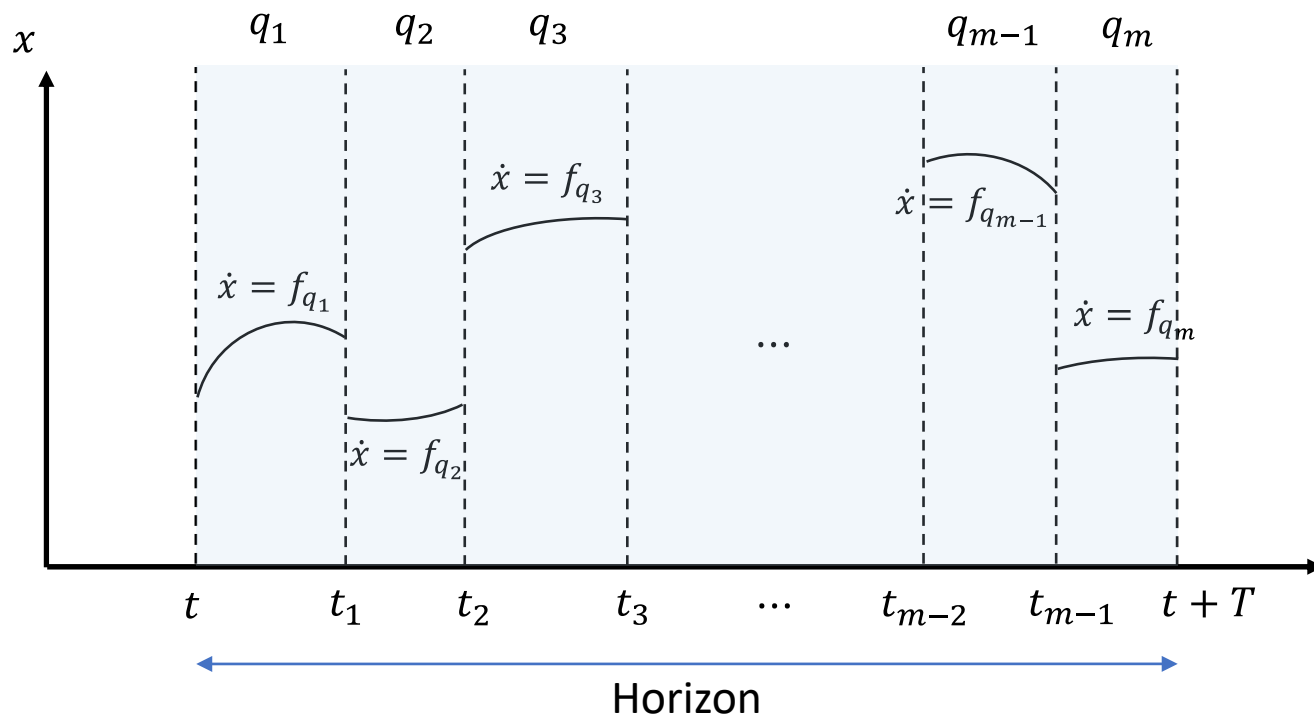


NMPC for Systems with SDS



OCP Formulation

- Switching sequence $\sigma = (q_1, \dots, q_m)$



OCP Formulation (for a given switching sequence)

Find the control input $u(t')$ ($t \leq t' \leq t + T$)

minimizing

$$J = \varphi_{q_m}(x(t + T)) + \int_{t_{m-1}}^{t+T} L_{q_m}(x(t'), u(t')) dt' + \sum_k^m \int_{t_{k-1}}^{t_k} L_{q_k}(x(t'), u(t')) dt' + \int_t^{t_1} L_{q_1}(x(t'), u(t')) dt'$$

subject to

$$\frac{d}{dt'} x(t') = f_{q_1}(x(t'), u(t')) \quad (t \leq t' < t_1)$$

$$\psi_{q_1, q_2}(x(t_1^-)) = 0, \quad x(t_1^+) = \gamma_{q_1, q_2}(x(t_1^-))$$

$$\begin{cases} \frac{d}{dt'} x(t') = f_{q_k}(x(t'), u(t')) & (t_{k-1} < t' \leq t_k) \\ \psi_{q_k, q_{k+1}}(x(t_k^-)) = 0, \quad x(t_k^+) = \gamma_{q_k, q_{k+1}}(x(t_k^-)) \end{cases}$$

$$\frac{d}{dt'} x(t') = f_{q_m}(x(t'), u(t')) \quad (t_{m-1} < t' \leq t + T)$$

OCP Formulation (for a given switching sequence)

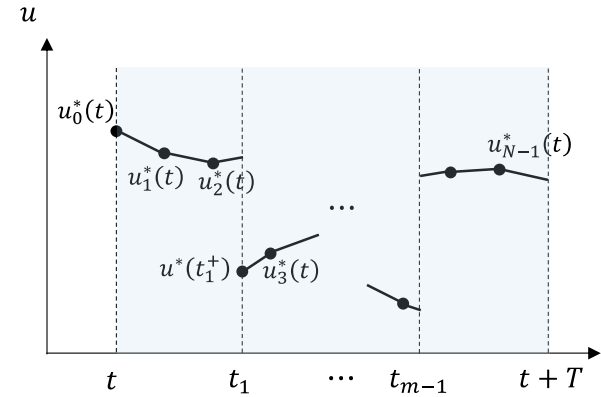
Find the variables to be determined

$$U(t) = \begin{bmatrix} u_0^*(t) \\ \vdots \\ u_{N-1}^*(t) \\ U_{q_1, q_2}(t) \\ \vdots \\ U_{q_{m-1}, q_m}(t) \end{bmatrix}$$

Discretized control input sequence on the horizon

$$U_{q_k, q_{k+1}}(t) = \begin{bmatrix} u^*(t_k^+) \\ v_{q_k, q_{k+1}}^*(t) \\ t_k \end{bmatrix}$$

variables with respect to switch from q_k to q_{k+1}



satisfying

$$F(U(t), x(t), t) := \begin{bmatrix} \nabla_u H_{q_1}(x_0^*(t), u_0^*(t), \lambda_1^*(t)) \\ \vdots \\ \nabla_u H_{q_m}(x_{N-1}^*(t), u_{N-1}^*(t), \lambda_N^*(t)) \\ F_{q_1, q_2}(U(t), x(t), t) \\ \vdots \\ F_{q_{m-1}, q_m}(U(t), x(t), t) \end{bmatrix} = 0,$$

State $x_i^*(t)$ and costate $\lambda_i^*(t)$ are functions of $U(t)$ and $x_0^*(0) = x(t)$ through ELE

Interior boundary conditions associated with switches

$\lambda \in \mathbb{R}^n$: Lagrange multiplier for $\dot{x} = f_q(x, u)$, $H_q(x, u, \lambda) = L_q(x, u) + \lambda^T f_q(x, u)$: Hamiltonian

C/GMRES Method

Nonlinear equation for $U(t)$

Find $U(t)$ satisfying $F(U(t), x(t), t) = 0$



Continuation method $\frac{d}{dt}F = -\zeta F$ ($\zeta > 0$)

Linear equation for $\dot{U}(t)$

Find $\dot{U}(t)$ satisfying $\frac{\partial F}{\partial U} \dot{U} = -\zeta F - \frac{\partial F}{\partial x} \dot{x} - \frac{\partial F}{\partial t}$



The GMRES method
(a fast Jacobian-free linear solver)

Update $U(t)$ on the basis of $\dot{U}(t)$, e.g., by

$$U(t + \Delta t) = U(t) + \dot{U}(t)\Delta t$$

**Only one linear equation
per update**

How to determine switching sequence?

Moving Switching Sequence Approach

Remove $U_{q_1, q_2}(t)$ from $U(t)$

A switch from q_1 to q_2 occurs on the actual system

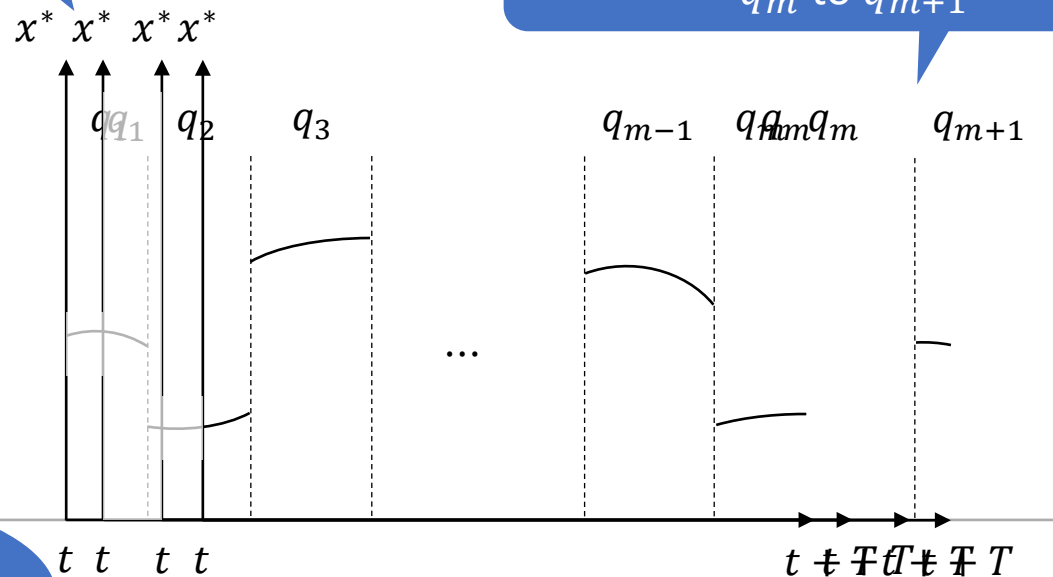
Modify the solution $U(t)$ taking the additional switch into account

Detect an additional switch from q_m to q_{m+1}

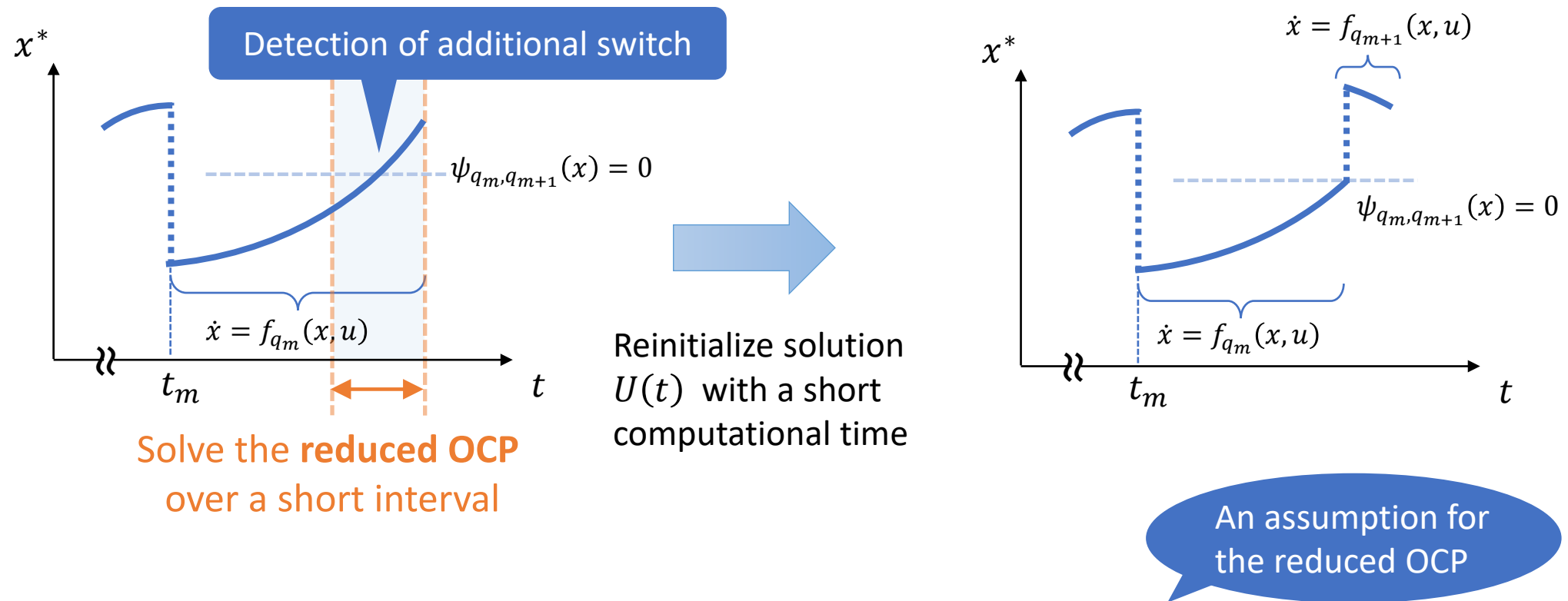
Switching sequence changes continuously

An assumption to justify this approach

Assumption 1 Switching sequence σ does not change except for the first and last elements



Reinitialization at Additional Switch



- **Assumption 2** Difference between the partial derivative of the terminal cost with respect to x before and after reinitialization is sufficiently small
- After reinitializing $U(t)$, we restart the continuation method for the OCP over the entire horizon

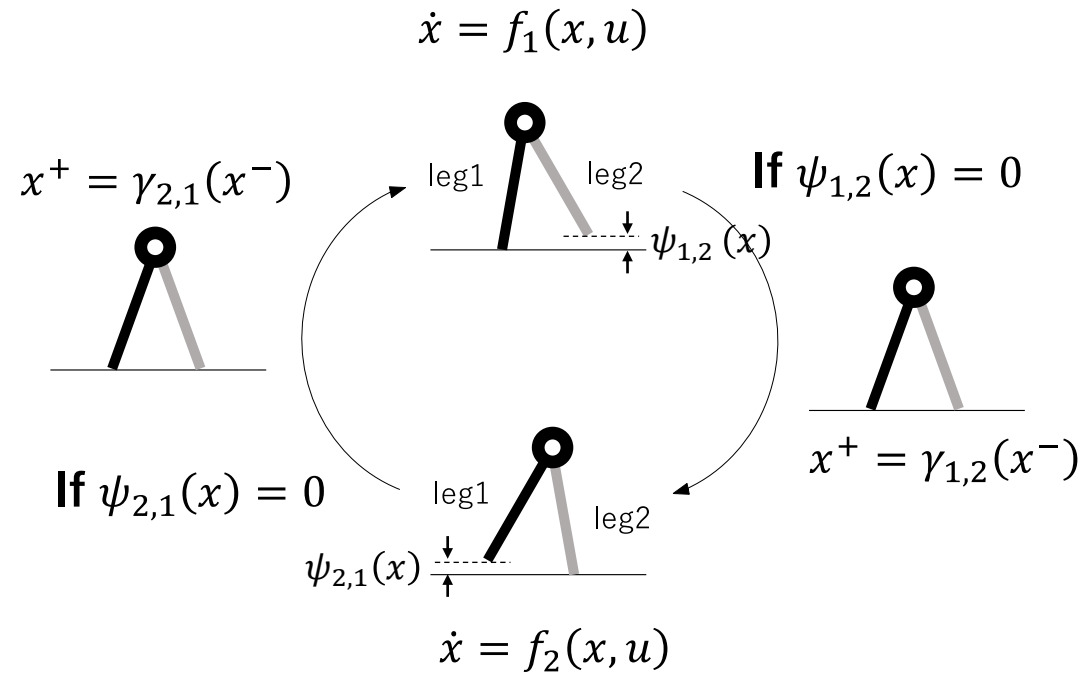
Summary of the Algorithm

At each time t

1. Compute the state trajectory on the horizon based on current $U(t)$ and σ
2. **If** an additional switch is detected **then**
3. Solve the reduced OCP to reinitialize $U(t)$
4. **End if**
5. Update $U(t + \Delta t)$ by the continuation method for the OCP over the entire horizon
6. **If** $t_1(t + \Delta t) < t + \Delta t$ **then**
7. Remove $U_{q_1, q_2}(t + \Delta t)$ from $U(t + \Delta t)$ and q_1 from σ
8. **End if**

Numerical Simulation

- Compass-like walking robot



- Performance index

$$L_i(x, u) = \frac{1}{2} a_1 (\dot{\theta}_i - v_{\text{ref}})^2 + \frac{1}{2} a_2 (\theta_1 + \theta_2)^2 + \frac{1}{2} r u^2, \quad i = 1, 2$$

$$\varphi_i(x) = 0, \quad i = 1, 2$$

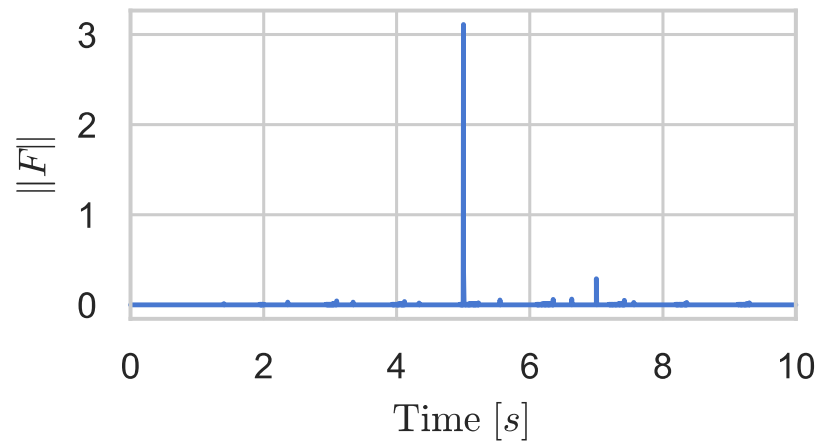
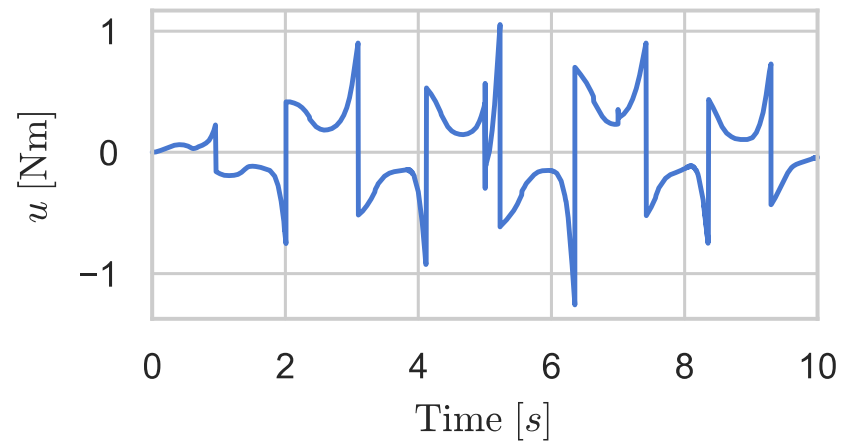
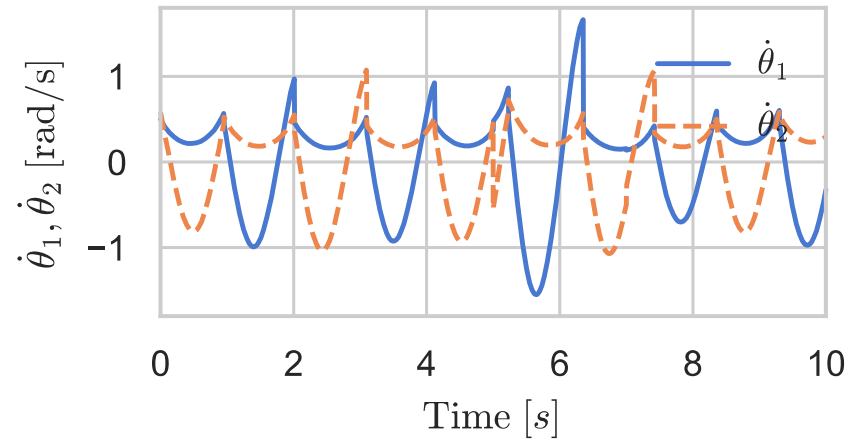
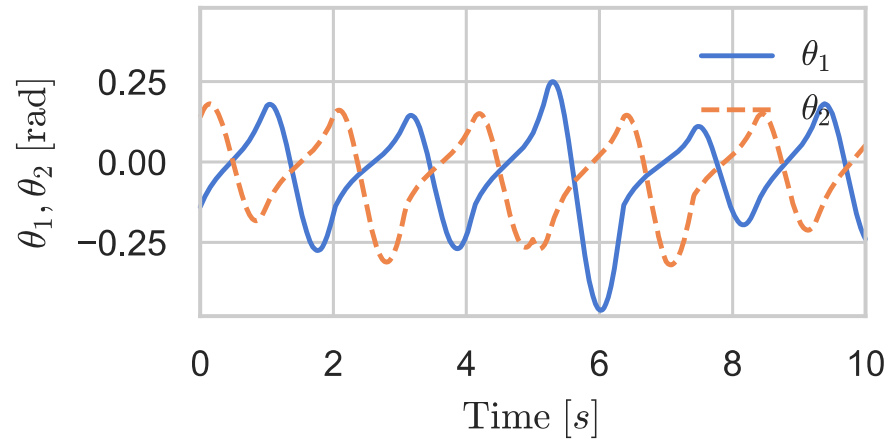
where $a_1, a_2, r, v_{\text{ref}} \in \mathbb{R}$

Numerical Simulation



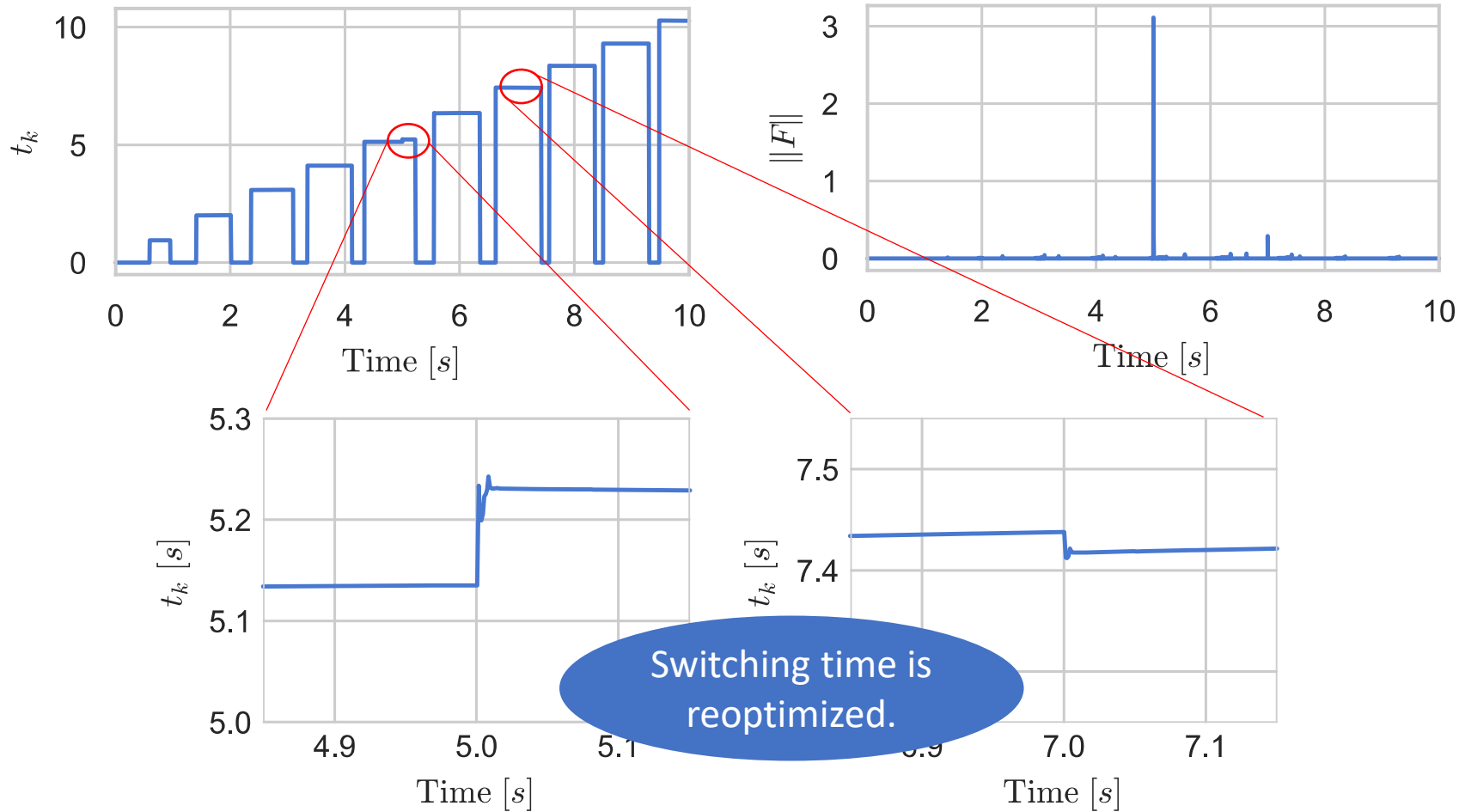
- Impulsive disturbances at 5 s and 7 s
- 0.14 ms per update

Numerical Simulation



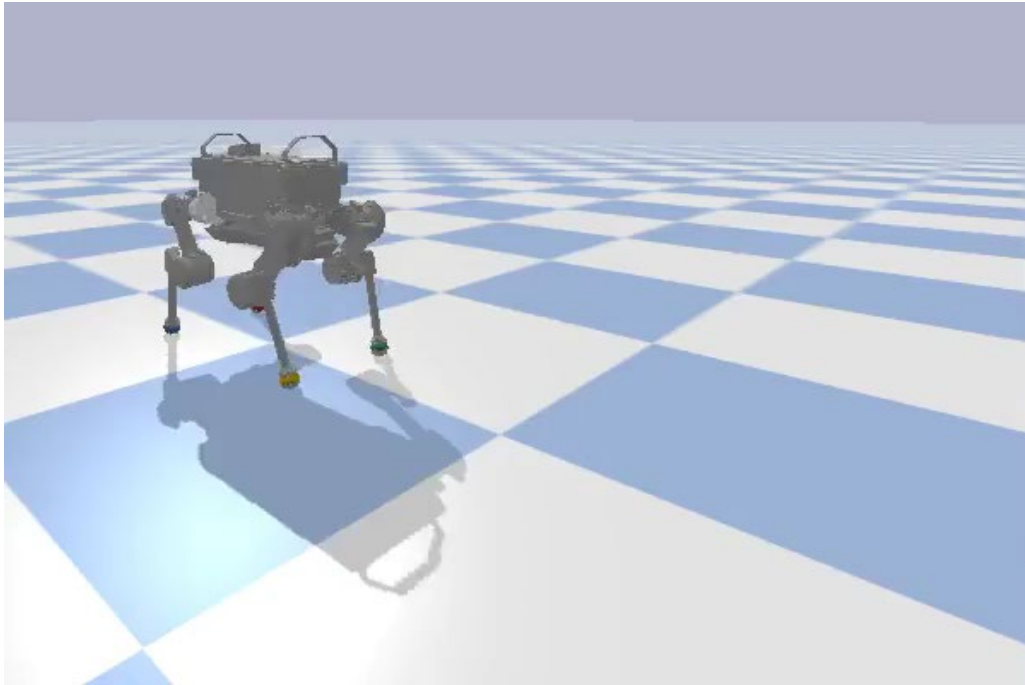
Impulsive disturbances
at 5 s and 7 s

Numerical Simulation

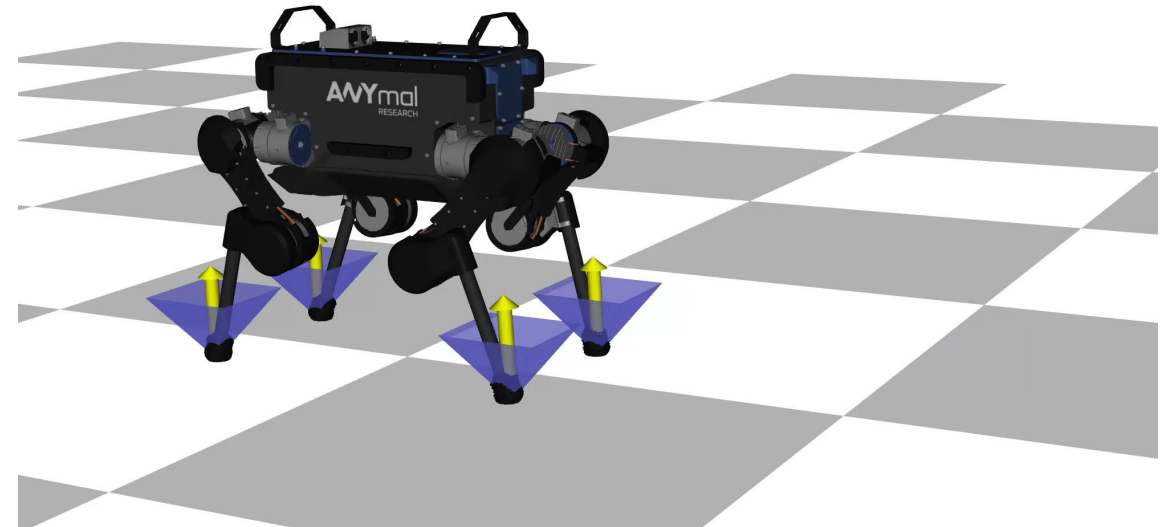


Impulsive disturbances
at 5 s and 7 s

More Complex System



Trotting by NMPC
(1.2ms per update of input)



Jump by OCP
(0.3 s in total for optimization)

Summary

- RTO algorithm for NMPC of nonlinear systems with **state-dependent switches and state jumps**
 - ✓ Based on an assumption that the **switching sequence is invariant except for the both ends of the horizon**
 - ✓ Solve a reduced OCP to reinitialize the solution when an additional switch is detected at the end of the horizon
- Succeeded in controlling a compass-like walking robot even when there are disturbances
- Ongoing work: Modifications, extensions, and software tools

Outline of Part 4

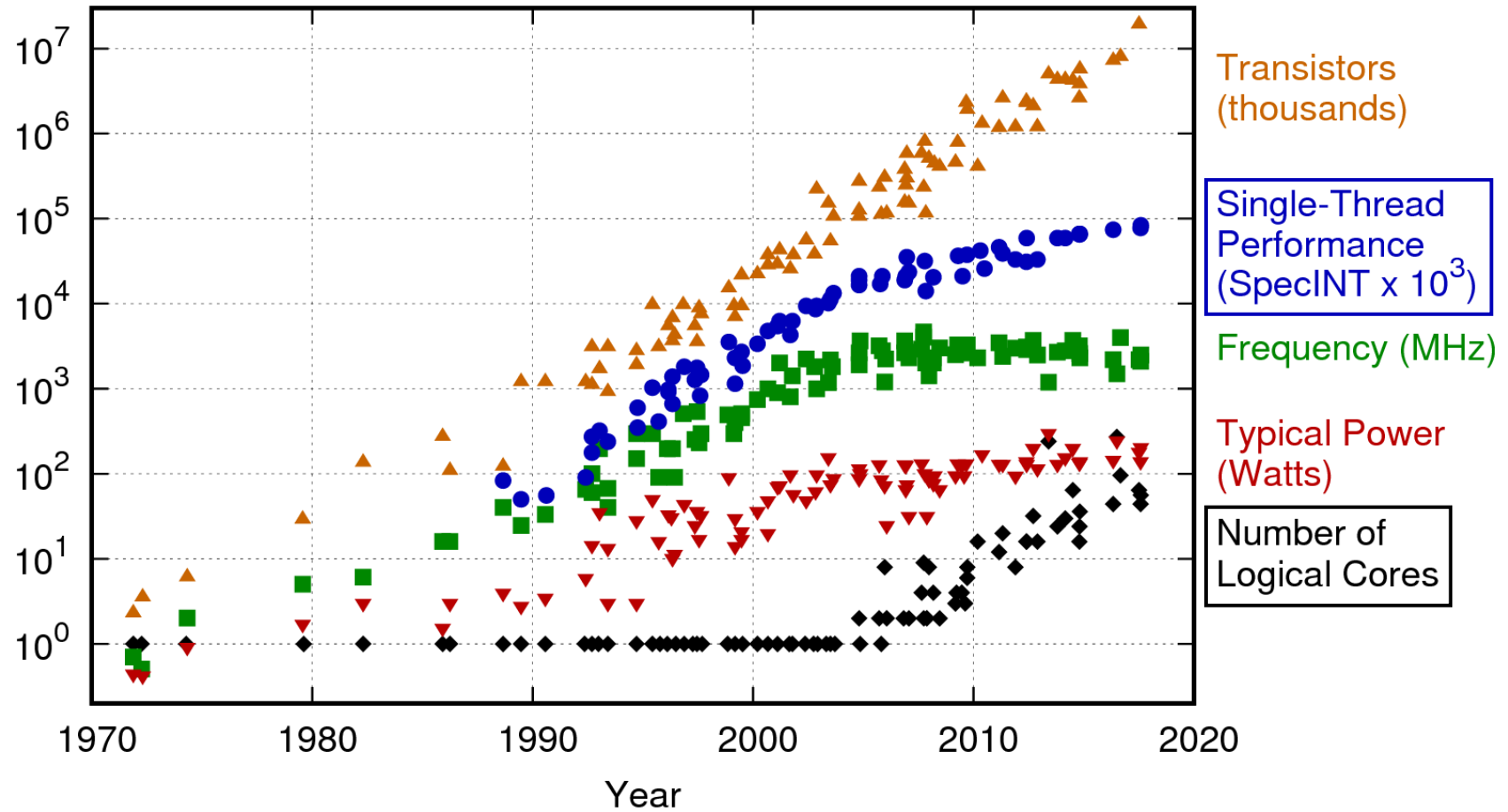
- NMPC with State-Dependent Switches and State Jumps
- **Parallel Algorithm for NMPC**

S. Katayama, M. Doi, T. Ohtsuka: A Moving Switching Sequence Approach for Nonlinear Model Predictive Control of Switched Systems with State-Dependent Switches and State Jumps; *Int. J. Robust & Nonlinear Contr.*, 30(2), 719-740 (2020)

H. Deng, T. Ohtsuka: A Parallel Newton-type Method for Nonlinear Model Predictive Control; *Automatica*, 109, paper 108560 (2019)

Motivation

42 Years of Microprocessor Trend Data



Parallel computing is a trend, however...

Motivation

Existing toolkits (algorithms) for NMPC:

- AutoGenU (C/GMRES)
- ACADO/acados (RTI/SQP)
- VIATOC (Gradient proj.)
- GRAMPC (Aug. Lagrangian & Gradient proj.)
- FORCES Pro (Interior-point)
-

Low degrees of parallelism

Speedup:
(Amdahl's law)

$$S(C) = \frac{1}{(1 - p) + \frac{p}{C}}$$

Non-parallelizable part $1 - p$

Parallelizable part p

↓ Parallelized on C cores

Non-parallelizable part $1 - p$

p/C

Motivation

$$\begin{aligned} & \min_{x(\cdot), u(\cdot)} \int_0^T L(u(\tau), x(\tau), p(\tau)) d\tau \\ \text{s.t.} \quad & x(0) = \bar{x}_0, \\ & \dot{x}(\tau) = f(u(\tau), x(\tau), p(\tau)), \quad \tau \in [0, T], \\ & C(u(\tau), x(\tau), p(\tau)) = 0, \quad \tau \in [0, T], \\ & G(u(\tau), x(\tau), p(\tau)) \geq 0, \quad \tau \in [0, T] \end{aligned}$$

Goal: Highly parallelizable algorithm & efficient toolkit

Key idea: Dividing the NMPC problem into subproblems along the prediction horizon

NMPC Problem

Inequality constraint elimination:

- To an equality constraint by introducing a dummy variable

$$G(u, x, p) \geq 0 \longrightarrow G(u, x, p) = v^2$$

- To an equality constraint and an additional cost (interior-point method)

$$G(u, x, p) \geq 0 \longrightarrow \begin{array}{l} G(u, x, p) = v \\ v \geq 0 \end{array} \longrightarrow \begin{array}{l} G(u, x, p) = v \\ -\rho \sum_j \log v_j \quad (\rho > 0) \end{array}$$

Equality-constraint NMPC problem:

$$\begin{array}{ll} \min_{x(\cdot), u(\cdot)} & \int_0^T L(u(\tau), x(\tau), p(\tau)) d\tau \\ \text{s.t.} & x(0) = \bar{x}_0, \\ & \dot{x}(\tau) = f(u(\tau), x(\tau), p(\tau)), \quad \tau \in [0, T], \\ & C(u(\tau), x(\tau), p(\tau)) = 0, \quad \tau \in [0, T] \end{array}$$

Couplings in NMPC Problem

- Parallelization is difficult because of the **couplings** introduced by the differential equation:

$$\dot{x}(\tau) = f(u(\tau), x(\tau), p(\tau)), \tau \in [0, T]$$

- Consider the optimization problem without the differential constraint:

$$\begin{aligned} \min_{x(\cdot), u(\cdot)} & \int_0^T L(u(\tau), x(\tau), p(\tau)) d\tau \\ \text{s.t.} & \quad x(0) = \bar{x}_0, \\ & \quad \dot{x}(\tau) = f(u(\tau), x(\tau), p(\tau)), \tau \in [0, T], \\ & \quad C(u(\tau), x(\tau), p(\tau)) = 0, \tau \in [0, T] \end{aligned}$$

It can be solved in parallel for each time stamp

Discretization for NMPC Problem

Generally, the discretized NMPC problem is solved:

Variables: $u(t), t \in [0, T]$
 $x(t), t \in [0, T]$

Variables: $U = (u_1, u_2, \dots, u_N)$
 $X = (x_0, x_1, x_2, \dots, x_N)$

$$\min_{x(\cdot), u(\cdot)} \int_0^T L(u(\tau), x(\tau), p(\tau)) d\tau$$

$$\sum_{i=1}^N L(u_i, x_i, p_i)$$

s.t. $x(0) = \bar{x}_0,$

$x_0 = \bar{x}_0$

$\dot{x}(\tau) = f(u(\tau), x(\tau), p(\tau)), \tau \in [0, T],$

Which disc. method should we use?

$C(u(\tau), x(\tau), p(\tau)) = 0, \tau \in [0, T]$

$C(u_i, x_i, p_i) = 0, i \in \{1, \dots, N\}$

Different discretization methods lead to different problem structures/couplings:

- Forward Euler method $x_i = x_{i-1} + hf(u, x_{i-1}, p)$
- Backward Euler method $x_i = x_{i-1} + hf(u, x_i, p)$
- Runge-Kutta method
-

Discretization method with a minimum coupling?

Discretization with Reduced Couplings

Reverse-time discretization method:

$$\dot{x}(\tau) = f(u(\tau), x(\tau), p(\tau)), \tau \in [0, T]$$

$$\downarrow$$
$$x_{i-1} + \mathcal{F}(u_i, x_i, p_i) = 0, i \in \{1, \dots, N\}$$

$$\text{e.g., backward Euler method: } x_{i-1} - x_i + \underbrace{hf(u_i, x_i, p_i)}_{\mathcal{F}(u_i, x_i, p_i)} = 0$$

Discretized problem (N -stage optimal control problem)

$$\min_{X, U} \sum_{i=1}^N L(u_i, x_i, p_i)$$

$$\text{s.t. } x_0 = \bar{x}_0,$$

$$x_{i-1} + \mathcal{F}(u_i, x_i, p_i) = 0, \quad i \in \{1, \dots, N\},$$

$$C(u_i, x_i, p_i) = 0, \quad i \in \{1, \dots, N\}$$

$$\text{where: } X = (x_0, x_1, x_2, \dots, x_N)$$

$$U = (u_1, u_2, \dots, u_N)$$

The couplings are “*reduced*” by using the reverse-time discretization method

KKT Conditions with Linear Couplings

- Karush-Kuhn-Tucker (KKT) conditions (necessary conditions for optimality):

Linear couplings

$$\begin{bmatrix} x_{i-1} + \mathcal{F}(u_i, x_i, p_i) \\ C(u_i, x_i, p_i) \\ \mathcal{H}_u^T(\lambda_i, \mu_i, u_i, x_i, p_i) \\ \lambda_{i+1} + \mathcal{H}_x^T(\lambda_i, \mu_i, u_i, x_i, p_i) \end{bmatrix} = 0, \quad i \in \{1, \dots, N\}$$

where: λ (**costate**) and μ : Lagrange multipliers

$$\mathcal{H}(\lambda, \mu, u, x, p) := L(u, x, p) + \lambda^T \mathcal{F}(u, x, p) + \mu^T C(u, x, p)$$

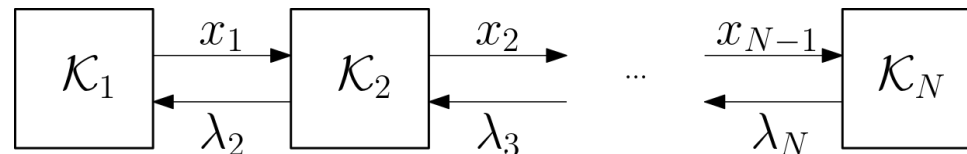
with initial and terminal conditions:

$$x_0 = \bar{x}_0 \quad \lambda_{N+1} = 0$$

- Compact form of the KKT conditions:

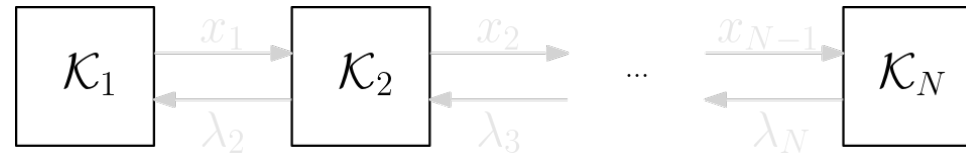
$$\mathcal{K}_i(x_{i-1}, s_i, \lambda_{i+1}) = 0, \quad i \in \{1, \dots, N\} \quad \text{or} \quad \mathcal{K}(S) = 0$$

$$s_i := (\lambda_i, \mu_i, u_i, x_i)$$



Parallel Method: Motivation

Assume that the optimal coupling variables are known in advance:



Then, the KKT conditions can be solved in parallel:

$$\mathcal{K}_i(x_{i-1}^*, s_i, \lambda_{i+1}^*) = 0, \quad i \in \{1, \dots, N\}$$

Single-stage problem

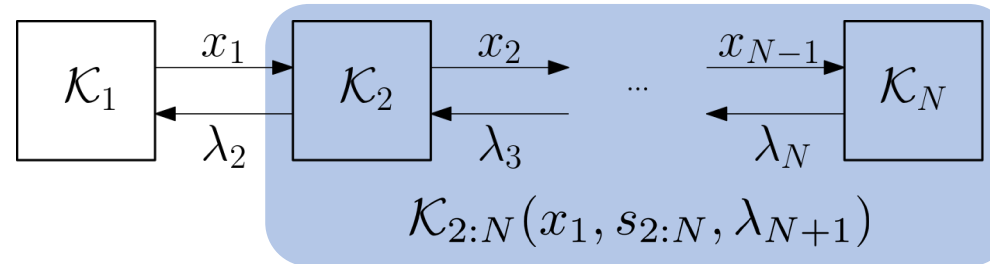
$$\begin{aligned} \min_{x_i, u_i} & L(u_i, x_i, p_i) + (\lambda_{i+1}^*)^T x_i \\ \text{s.t.} & x_{i-1}^* + \mathcal{F}(u_i, x_i, p_i) = 0, \\ & C(u_i, x_i, p_i) = 0 \end{aligned}$$

However, it is impossible to know the coupling variables (x_{i-1}^* and λ_{i+1}^*) in advance!

Can we estimate them?

Linearity in Costate

Consider the estimation of λ_2^* :



- Solving $\mathcal{K}_{2:N}$ using Newton's method (k -th iteration):

$$s_{2:N}^{k+1}(x_1) = s_{2:N}^k - \underbrace{\left(\frac{\partial \mathcal{K}_{2:N}}{\partial s_{2:N}} \bigg|_{(x_1, s_{2:N}^k, \lambda_{N+1})} \right)^{-1}}_{\text{Independent of } x_1} \mathcal{K}_{2:N}(x_1, s_{2:N}^k, \lambda_{N+1})$$

\downarrow
 $\mathcal{K}_{2:N}(x_1^k, s_{2:N}^k, \lambda_{N+1}) + [x_1 - x_1^k \ 0]$

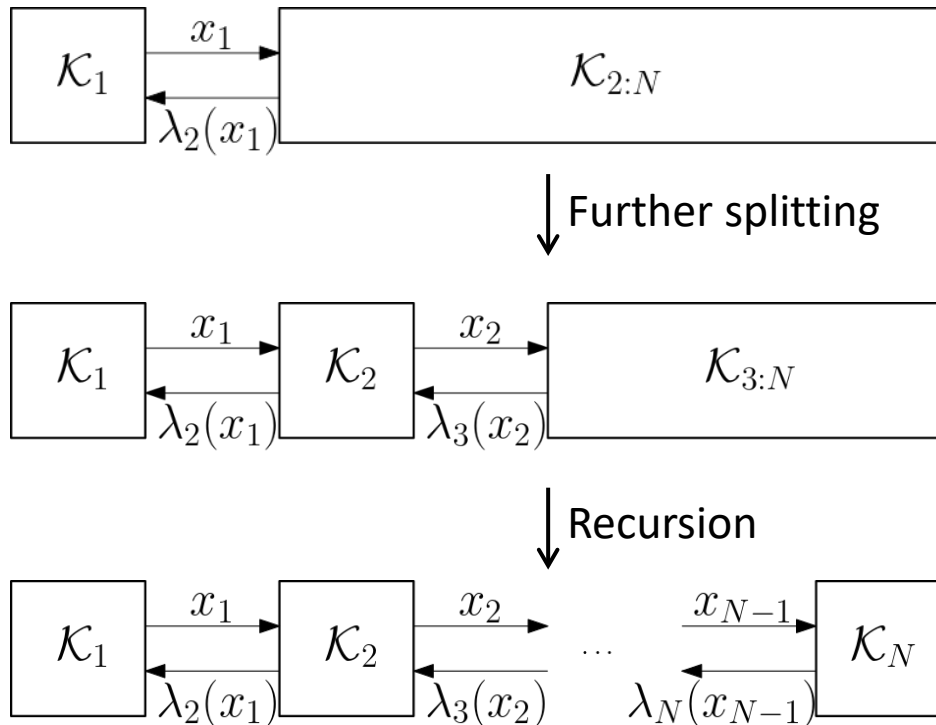
Linear coupling of x_1

- Extracting $\lambda_2(x_1)$ from $s_{2:N}^{k+1}(x_1)$:

$$\lambda_2^{k+1}(x_1) = \lambda_2^k - \Lambda_2^k(x_1 - x_1^k) - d_{\lambda_2}^k \quad (\text{estimation of } \lambda_2^*)$$

Backward Correction Method

Estimation of the costate variables:



Backward correction method

- Estimating $\lambda_{N:1}^*$ (backward)
- Solving $\mathcal{K}_{1:N}$ (forward)

Backward correction method
= Newton's method

$$S^{k+1} = S^k - \nabla \mathcal{K}(S^k)^{-1} \mathcal{K}(S^k)$$

$$\lambda_i^{k+1}(x_{i-1}) = \lambda_i^k - \Lambda_i^k(x_{i-1} - x_{i-1}^k) - d_{\lambda_i}^k$$

Then, we can solve \mathcal{K} in a forward manner

Backward Correction Method

Backward correction method (structure-exploiting Newton's method):

Algorithm 1 k -th iteration

Input: \bar{x}_0, S^k

Output: S^{k+1}

for $i = N$ to 1 **do**

$$\mathcal{K}_i^k \leftarrow \mathcal{K}_i(x_{i-1}^k, s_i^k, \lambda_{i+1}^k)$$

KKT & Jacobian evaluations

$$J_i^k \leftarrow \nabla_{s_i} \mathcal{K}_i(x_{i-1}^k, s_i^k, \lambda_{i+1}^k)$$

$$H_i^k \leftarrow \left(J_i^k - \begin{bmatrix} 0 & 0 \\ 0 & \Lambda_{i+1}^k \end{bmatrix} \right)^{-1}$$

Costate λ estimation (backward)

$$\lambda_i^{k+1}(x_{i-1}) = \lambda_i^k - \Lambda_i^k(x_{i-1} - x_{i-1}^k) - d_{\lambda_i}^k$$

$$\Lambda_i^k \leftarrow \begin{bmatrix} 0 & 0 \\ I_{n_x} & 0 \end{bmatrix} H_i^k \begin{bmatrix} 0 & I_{n_x} \\ 0 & 0 \end{bmatrix}$$

$$\begin{bmatrix} 0 \\ d_{\lambda_i}^k \end{bmatrix} \leftarrow \begin{bmatrix} 0 & 0 \\ I_{n_x} & 0 \end{bmatrix} H_i^k \left(\mathcal{K}_i^k - \begin{bmatrix} 0 \\ d_{\lambda_{i+1}}^k \end{bmatrix} \right)$$

Most computationally expensive part & in **serial**

end for

for $i = 1$ to N **do**

$$s_i^{k+1} \leftarrow s_i^k - H_i^k \left(\mathcal{K}_i^k - \begin{bmatrix} 0 \\ d_{\lambda_{i+1}}^k \end{bmatrix} + \begin{bmatrix} \Delta x_{i-1}^k \\ 0 \end{bmatrix} \right)$$

Iteration (forward)

end for

How can we parallelize this algorithm?

Approximation for Parallelization

- Estimation of λ_{i+1}^* in the backward correction method:

$$\lambda_{i+1}^{k+1}(x_i) = \lambda_{i+1}^k - \Lambda_{i+1}^k(x_i - x_i^k) - d_{\lambda_{i+1}}^k$$

$$H_i^k \leftarrow \left(J_i^k - \begin{bmatrix} 0 & 0 \\ 0 & \Lambda_{i+1}^k \end{bmatrix} \right)^{-1}$$

- Accurate estimation
- Time-consuming (**recursion**)

- A coarse estimation of λ_{i+1}^* based on the previous iteration's information:

$$\lambda_{i+1}^{k+1}(x_i) = \lambda_{i+1}^k - \Lambda_{i+1}^{k-1}(x_i - x_i^k) - d_{\lambda_{i+1}}^k$$

$$H_i^k \leftarrow \left(J_i^k - \begin{bmatrix} 0 & 0 \\ 0 & \Lambda_{i+1}^{k-1} \end{bmatrix} \right)^{-1}$$

- Approximate estimation
- Fast (in **parallel**)

Parallel Newton-type Method

Parallel algorithm:

Algorithm 1 k -th iteration

Input: $\bar{x}_0, S^k, \Lambda_{1:N}^{k-1}$

Output: $S^{k+1}, \Lambda_{1:N}^k$

for $i = 1$ to N **do in parallel**

$$\mathcal{K}_i^k \leftarrow \mathcal{K}_i(x_{i-1}^k, s_i^k, \lambda_{i+1}^k)$$

$$J_i^k \leftarrow \nabla_{s_i} \mathcal{K}_i(x_{i-1}^k, s_i^k, \lambda_{i+1}^k)$$

$$H_i^k \leftarrow \left(J_i^k - \begin{bmatrix} 0 & 0 \\ 0 & \Lambda_{i+1}^{k-1} \end{bmatrix} \right)^{-1}$$

$$\Lambda_i^k \leftarrow \begin{bmatrix} 0 & 0 \\ I_{n_x} & 0 \end{bmatrix} H_i^k \begin{bmatrix} 0 & I_{n_x} \\ 0 & 0 \end{bmatrix}$$

end for

for $i = N$ to 1 **do**

$$\begin{bmatrix} 0 \\ d_{\lambda_i}^k \end{bmatrix} \leftarrow \begin{bmatrix} 0 & 0 \\ I_{n_x} & 0 \end{bmatrix} H_i^k \left(\mathcal{K}_i^k - \begin{bmatrix} 0 \\ d_{\lambda_{i+1}}^k \end{bmatrix} \right)$$

end for

for $i = 1$ to N **do**

$$s_i^{k+1} \leftarrow s_i^k - H_i^k \left(\mathcal{K}_i^k - \begin{bmatrix} 0 \\ d_{\lambda_{i+1}}^k \end{bmatrix} + \begin{bmatrix} \Delta x_{i-1}^k \\ 0 \end{bmatrix} \right)$$

end for

- KKT & Jacobian evaluations
- Matrix factorization

- In parallel
- Degree of parallelism N

- In serial
- $2N$ matrix-vector multiplications

ParNMPC Toolkit

ParNMPC (Parallel NMPC)

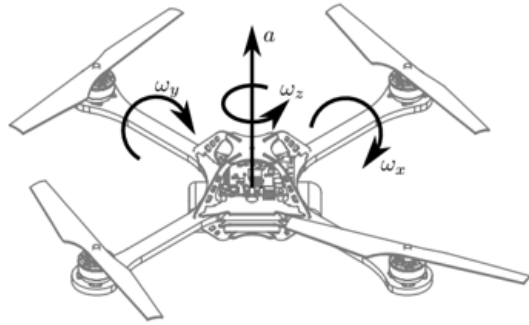
- Open source
- Symbolic problem formulation
- Primal-dual interior-point method
- Automatic parallel code generation
 - C/C++
 - lib, dll, mex, exe

Options (version 1903)

- Hessian approximation
- High-order discretization
- Degree of parallelism configuration
- NLP techniques
 - Regularization
 - Line search
 - Barrier strategy
 - Warm start strategy
 -

- Fast
- Numerically robust
- Easy-to-use

Numerical Experiment: Quadrotor



$$\ddot{X} = a(\cos \gamma \sin \beta \cos \alpha + \sin \gamma \sin \alpha)$$

$$\ddot{Y} = a(\cos \gamma \sin \beta \sin \alpha - \sin \gamma \cos \alpha)$$

$$\ddot{Z} = a \cos \gamma \cos \beta - g$$

$$\dot{\gamma} = (\omega_X \cos \gamma + \omega_Y \sin \gamma) / \cos \beta$$

$$\dot{\beta} = -\omega_X \sin \gamma + \omega_Y \cos \gamma$$

$$\dot{\alpha} = \omega_X \cos \gamma \tan \beta + \omega_Y \sin \gamma \tan \beta + \omega_Z$$

Quadrotor position control

- 9 states, 4 inputs
- $T = 0.5$ s, $N = 24$
- Quadratic cost function:

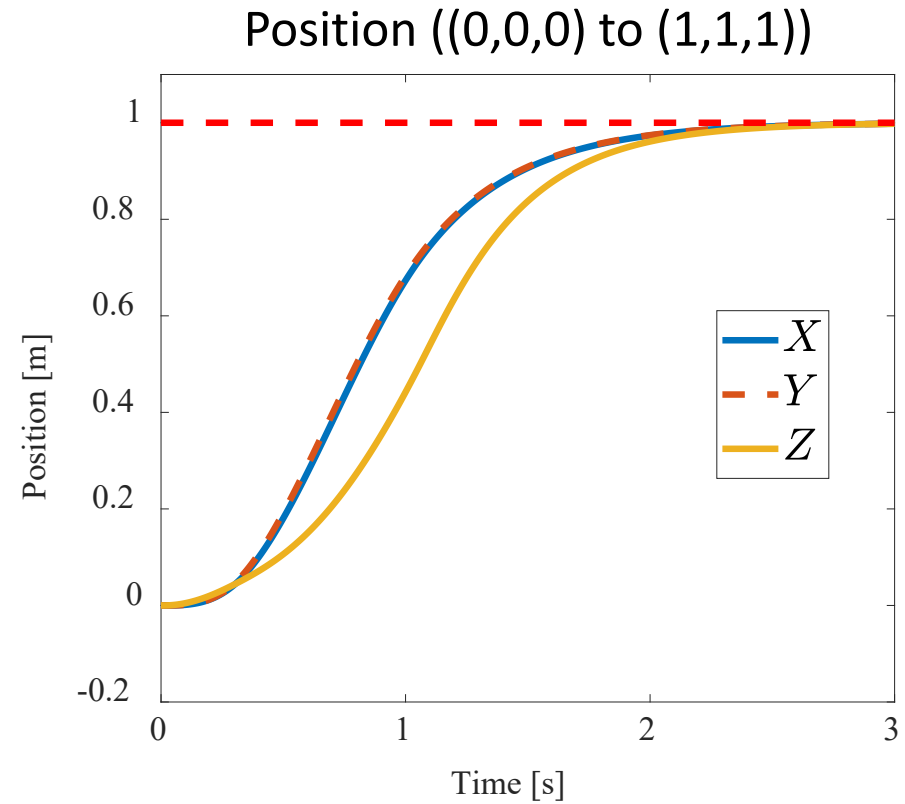
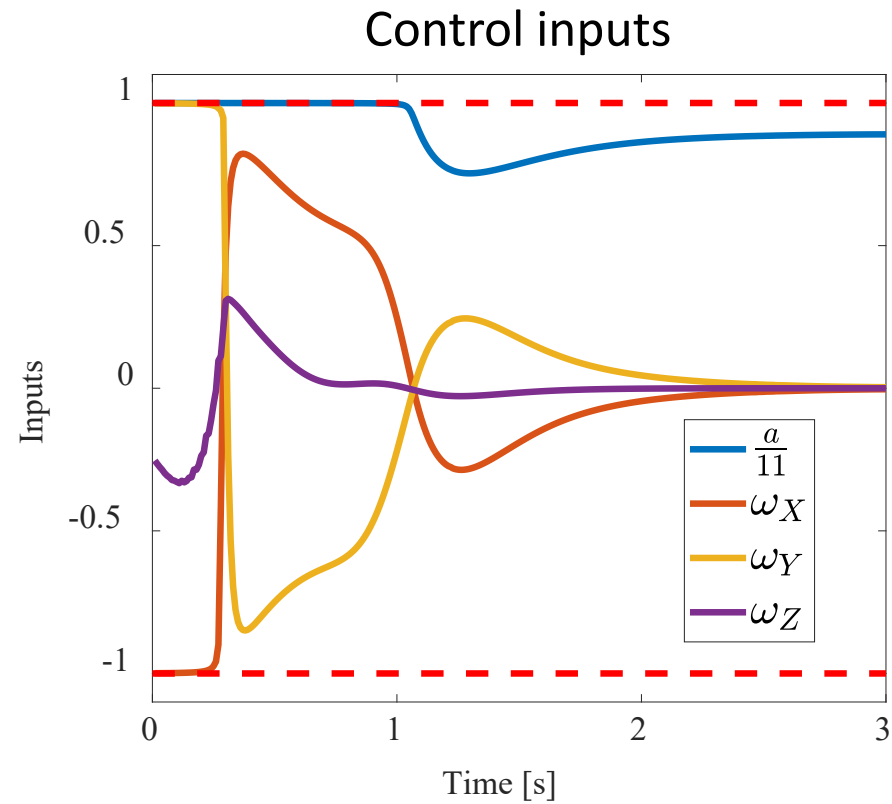
$$L(u, x, p) := \frac{1}{2} (\|x - x_r\|_Q^2 + \|u - u_r\|_R^2)$$

- Input constraints:

$$\begin{bmatrix} 0 \text{ m/s}^2 \\ -1 \text{ rad/s} \\ -1 \text{ rad/s} \\ -1 \text{ rad/s} \end{bmatrix} \leq u \leq \begin{bmatrix} 11 \text{ m/s}^2 \\ 1 \text{ rad/s} \\ 1 \text{ rad/s} \\ 1 \text{ rad/s} \end{bmatrix}$$

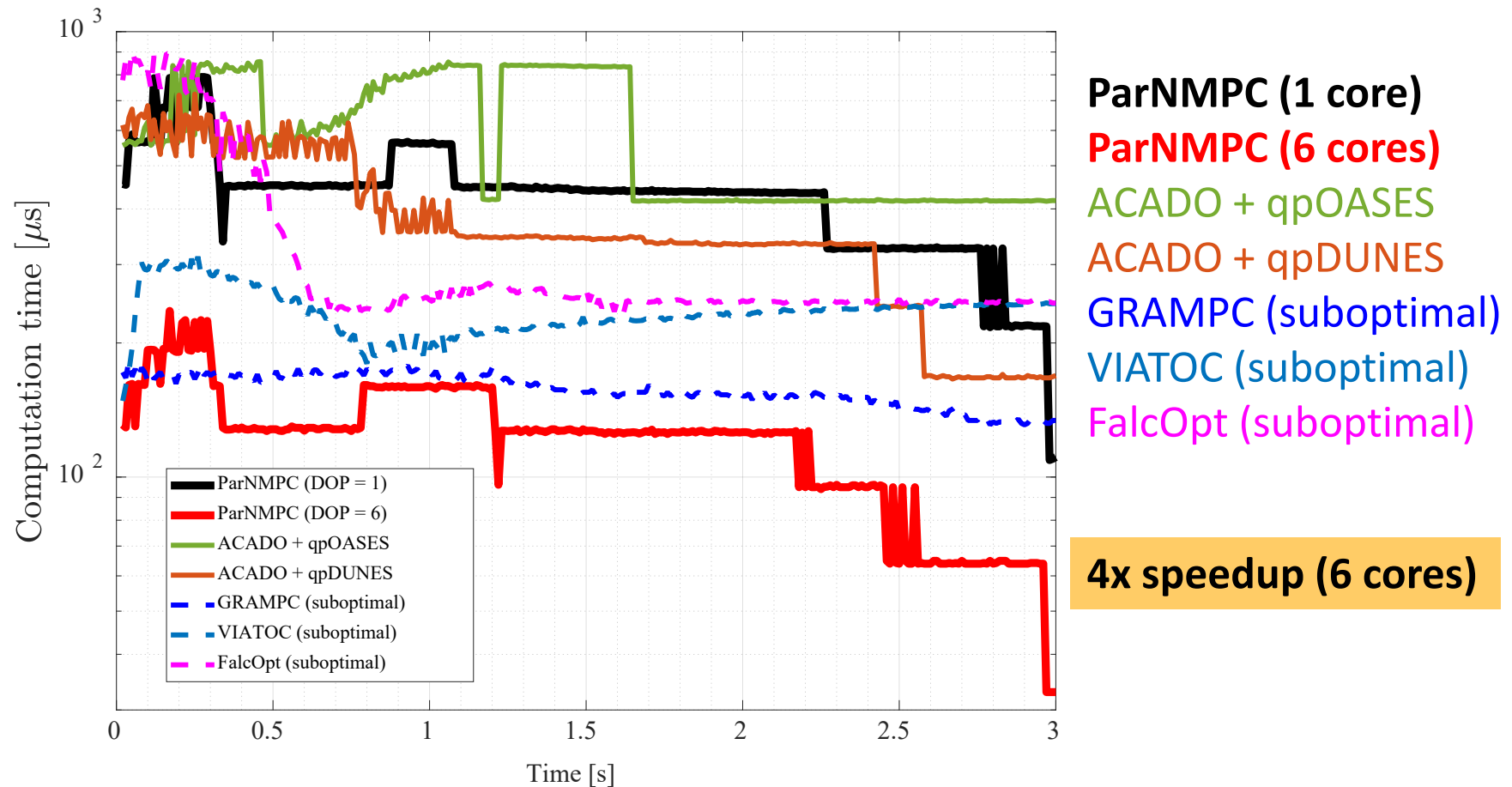
- Intel Core i9-8950HK @2.9 GHz, 6 cores

Numerical Experiment: Quadrotor



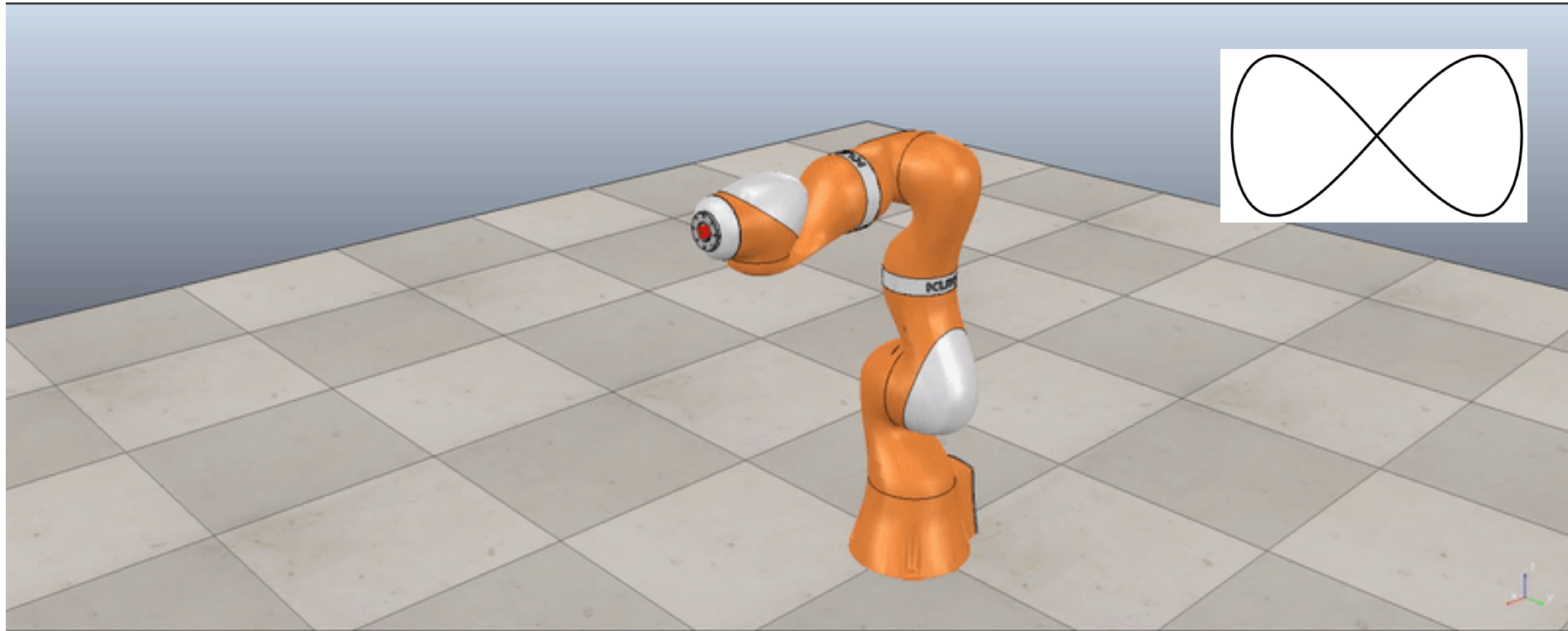
Numerical Experiment: Quadrotor

Comparison with state-of-the-art solvers



Numerical Experiment: Robot Manipulator

Path tracking control of a 7 DOF robot manipulator:



Constraints:

- Input torque
- Angular velocity

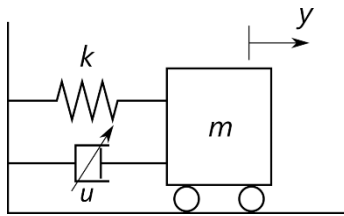
Results ($T = 1$ s, $N = 18$)

- **240 us/iteration**
- **5.6x speedup** (6 cores)

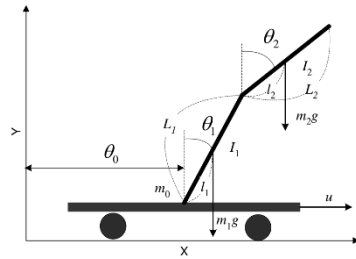
1 kHz real-time NMPC

Other Applications

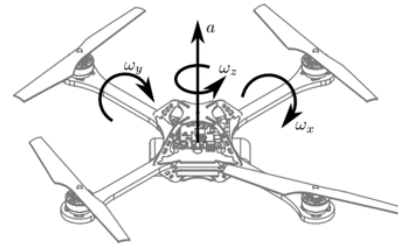
Successful applications of ParNMPC:



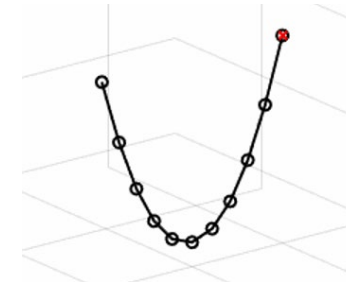
Semi-active damper



Double inverted pendulum



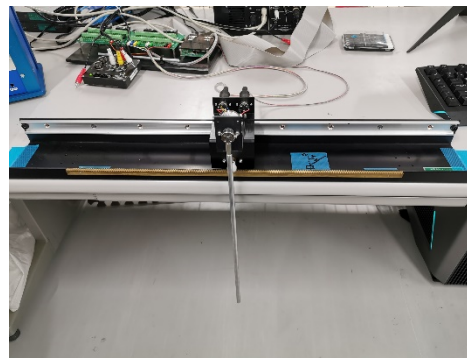
Quadrotor



Chain of masses



Helicopter



Inverted pendulum on a cart



7 DOF robot manipulator

...

Summary

- **Parallel Newton-type method**
 - ✓ Highly parallelizable
 - ✓ Fast rate of convergence
 - ✓ Applicable to general NMPC problems
- **ParNMPC (code generation tool)**
 - ✓ Easy-to-use
 - ✓ Automatic parallel code generation
 - ✓ Large number of applications
- **Future directions**
 - ✓ Reliable parallel computing
 - ✓ FPGA/GPU implementation

Conclusions

- Control systems are everywhere and provide motivations and opportunities for artificial intelligence.
- **Model predictive control (MPC) by real-time optimization** is very powerful as long as mathematical models are available.
- Most control systems have some mathematical models, but it is often difficult to find **appropriate models for control**. It is also often difficult to define **appropriate performance indices**.
- Machine learning would be particularly useful for fine tuning of mathematical models and/or performance indices.
- Application of artificial intelligence would be expected to the entire process of analysis and design of control systems.